

VHDL

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0.1 Einführung

VHDL (VHSIC (Very High Speed Integrated Circuits) Hardware Description Language) ist in Europa die verbreitetste Hardware-Beschreibungssprache. Daneben gibt es VERILOG. Ursprünglich wurde sie entwickelt, um Testumgebungen für die Simulation integrierter Schaltungen zu entwickeln. Daher ist VHDL eine relativ komplexe Programmiersprache, deren Konstrukte nicht zwangsläufig synthetisierbar sind. Das führt bereits zum üblichen Ablauf der Hardwareentwicklung: Nach der Festlegung der Funktionalität wird diese mittels VHDL beschrieben. Der entstandene "Source-Code" wird kompiliert und anschließend simuliert. Nachdem die einwandfreie Funktion sichergestellt ist wird der VHDL-Code direkt mit Hilfe eines Syntheseprogramms in eine Gatternetzliste umgesetzt. Dazu benötigt das Synthesewerkzeug eine von der Zielhardware abhängige Elementbibliothek, die der Chiphersteller zur Verfügung stellt.

0.1.1 Zum Wesen von VHDL

Im Innersten besteht jede digitale Hardwareschaltung aus kombinatorischer Logik und speichernden Elementen.

Unter **Kombinatorik** versteht man *NICHT/UND/ODER/Exclusiv-ODER*-Gatter und deren Kombinationen. Also jede Art von Verknüpfungen von einem oder mehreren Eingängen zu einem oder mehreren Ausgängen. Eine Änderung eines Eingangs bewirkt eine unmittelbare Wirkung auf den Ausgang. Eine wirkliche Hardwareschaltung benötigt dazu jedoch

geringe Laufzeiten. Diese Laufzeiten kombinatorischer Logik werden aber in einer reinen *RTL*- Beschreibung nicht im VHDL-Code modelliert, obwohl dies an sich möglich wäre.

Speichernde Elemente sind *Flipflops* oder *Latches*. Latches sollte man als Zustandsspeicher in synchronen Schaltungen vermeiden. Flipflops sind Grundelemente, die einen Dateneingang und einen Takteingang haben. Der Ausgang übernimmt üblicherweise den Zustand des Eingangs mit der steigenden Taktflanke. Es gibt auch Flipflops, die mit der fallenden Flanke schalten. Zusätzlich können Flipflops auch einen asynchronen Reset-Eingang haben. Dieser setzt im Normalfall zu Beginn, nach Anlegen der Versorgungsspannung, das Flipflop in den Grundzustand (Ausgang hat den Pegel „0“).

Um diese zwei Arten der realen Hardware nachzubilden, ist die grundsätzliche Denkweise für ein VHDL-Programm deutlich anders, als es für den seriellen Ablauf beispielsweise eines C-Programms nötig ist. VHDL ist im Grunde eine Aneinanderreihung von **Prozessen**, die quasi simultan abgearbeitet werden. Der VHDL-Simulator wird zwar den Code in irgendeiner Weise in eine serielle Software umwandeln. Die Wirkungsweise der Prozesse ist jedoch so, als würden sie wirklich völlig gleichzeitig bearbeitet.

Zu welchem Zeitpunkt ein Prozess abgearbeitet wird, bestimmt die **Sensitivity-List**. Dies ist im Prozess-Header eine Liste von vereinbarten Signalen. Ändert eines dieser Signale den Wert, so wird der Prozess angestoßen. Führt dieses zu einer Änderung eines Ausgangs und ist dieser wiederum in der Sensitivity-List eines anderen Prozesses, so wird auch dieser mit dem neuen Wert des Signals angestoßen.

Man unterscheidet zwei Typen von Prozessen: Kombinatorische und synchron getaktete Prozesse, analog zu der eingangs besprochenen realen Hardware. Kombinatorische Prozesse haben in der Sensitivity-List alle Eingangssignale und beschreiben im Inneren deren Verknüpfung. Synchron getaktete Prozesse haben in der Sensitivity-List nur „reset“ und „clock“. Im Inneren wird beschrieben, welches Signal oder auch welche Verknüpfung von Signalen zur Taktflanke am Ausgang übernommen werden soll.

Beispiel für einen kombinatorischen Prozess:

```

procname: process(a,b,c)
begin
  x <= (a and b) or c;
end process;

```

-- Ausgang "x" ist eine Verknüpfung von "a","b","c"

Beispiel für einen synchron getakteten Prozess (und asynchronem, low-aktivem Reset):

```

procname: process(nres,clk)
begin
  if (nres='0') then
    q <= '0';
  elsif (clk'event and clk='1') then
    q <= x;
  end if;
end process;

```

-- FlipFlop mit Ausgang "q" schaltet mit steigender "clk"-Flanke und übernimmt den Wert von "x" -- x stammt aus kombinatorischem Prozess!!

Zur Beschreibung kombinatorischer Vorgänge gibt es die aus vielen Sprachen bekannte Konstrukte:

```
n: process(a,b)
begin
  x <= a and b;    -- Und-Verknüpfung von "a" und "b"
end;
```

--dazu gleichwertig:

```
n: process(a,b)
begin
  if (a='1') and (b='1') then
    x <= '1';
  else
    x <= '0';
  end if;
end;
```

--dazu gleichwertig:

```
n: process(a,b)
variable ab : std_logic_vector(1 downto 0);
begin
  ab(0) := a;
  ab(1) := b;
  case ab is
    when "00" => x <= '0';
    when "01" => x <= '0';
    when "10" => x <= '0';
    when "11" => x <= '1';
    when others => null;
  end case;
end;
```

Im Weiteren baut VHDL einen begrenzten Rahmen von logischen Elementen als ein Bauelement zusammen, das wiederum mit der diskreten Schaltungstechnik vergleichbar ist. So ist es ähnlich wie in der altbekannten TTL-Schaltungstechnik, dass ein solches Bauteil Eingänge, Ausgänge und ein Innenleben hat. VHDL definiert dieses Element oder diesen Block mit seinen "Pins" in der "**Entity**". Die "**Architecture**" beschreibt dann mit den oben gezeigten Prozessen das Innenleben.

Formal:

```
entity 74LS00 is
  port (i0,i1 : in bit;
        i2,i3 : in bit;    -- Eingänge
        o0   : out bit;    -- Ausgang
        o1   : out bit);
end 74LS00;
```

```
architecture behv of 74LS00 is
begin
signal z : bit;      -- Vereinbarung internen Signale

comb_1: process(i0,i1,i2,i3)
begin
  o0 <= not(i0 and i1);  -- nand nr.1
  z  <= not(i2 and i3);  -- nand nr.2
end comb_1;

comb_2: process(z)
begin
  o1 <= z;
end comb_2;

end behv;
```

Man sieht: In der "entity" wird eine Portliste vereinbart, die alle nach außen führenden Signale beinhalten muss. Interne Signale werden wie oben gezeigt vereinbart. Üblicherweise führen diese nach der Synthese zu realen "Drähten" in der Schaltung, können aber auch, wie in diesem Fall "z", wegoptimiert werden. Besonderheit: Signale, die den Block verlassen, können *nicht* in der "architecture" verschaltet werden. Das heißt, alle in der entity als "out" deklarierten Signale können nirgends in der "sensitivity list" eines Prozesses oder als Zuweisungswert erscheinen. Sie sollen ebenfalls nur in einem einzigen Prozess des VHDL-Files zugewiesen werden, so wie ein "Draht" ebenfalls zu einem Zeitpunkt immer nur einen Pegel führen kann.

Ein File mit einer Entity und einer Architecture ist bereits eine kompilierbare, vollständige VHDL-Komponente. Ähnlich wie auf einer Platine können auch mehrere VHDL-Bauteile miteinander quasi verdrahtet werden. In diesem Fall wird in einem anderen oder auch gleichen VHDL-File diese Komponente als "**component**" eingebunden.

Beispiel einer "Component"-Deklaration:

```
component 74LS00
  port (i0,i1,i2,i3 : in bit;
        o0,o1      : out bit);
end component;
```

Beispiel für eine "Component"-Instanziierung:

```
architecture behv of test is
  -- Signal-Deklarierungen:
  signal a,b,c,d : bit;

  -- Komponenten-Deklarierungen:

  component adder
    port(a,b      : in bit;
         sum,carry : out bit);
  end component;

begin
```

```

    teil_a: adder port map (a,b,c,d); -- Komponente "adder" hat den Namen
    "teil_a"
        -- a,b,c,d ist angeschlossen

end behv;

```

0.2 Basis-Konstrukte

Nach dieser kurzen Einführung über die grundsätzlichen Gedanken von VHDL sollen im Folgenden in alphabetischer Reihenfolge die sprachlichen Basis-Konstrukte beschrieben werden:

0.2.1 aggregates

Ein Aggregat ist ein Klammerausdruck, der mehrere Einzelemente zu einem Vektor zusammenfasst, wobei die Elemente durch Kommata getrennt werden.

(wert_1,wert_2,...)

(element_1 => wert_1,element_2 => wert_2,...)

Beispiele:

```

signal databus : bit_vector(3 downto 0);
signal d1,d2,d3,d4 : bit;
...
databus <= (d1,d2,d3,d4);

```

identisch mit:

```

databus(3) <= d1;
databus(2) <= d2;
databus(1) <= d3;
databus(0) <= d4;

```

```

--
type zustand is (idle,run,warte,aktion); -- enumerated type
signal state : zustand;
--

type packet is record
    flag    : std_logic;
    nummer  : integer range 0 to 7;
    daten   : std_logic_vector(3 downto 0);
end record;
signal x : packet;
...
x <= ('1',3,"0011");
--

type vierbit is array(3 downto 0) of std_ulogic;
type speicher is array(0 to 7) of vierbit;

variable xmem : speicher := (others=>'0'); -- mit '0' vorbelegen
--

```

```
variable dbus : std_logic_vector(15 downto 0) := (others=>'Z');
```

0.2.2 alias

Ein Alias bezeichnet einen Teil eines Signals.

```
alias 'alias_name' : 'alias_type'(range) is 'signal_name'(range);
```

Beispiel:

```
signal daten_in : bit_vector(11 downto 0);  
alias opcode : bit_vector(3 downto 0) is daten_in(3 downto 0);  
alias daten : bit_vector(7 downto 0) is daten_in(11 downto 4);
```

0.2.3 architecture

Eine Designeinheit in VHDL, die das Verhalten oder die Struktur einer Entity beschreibt.

```
architecture 'architecture_name' of 'entity_name' is  
    declarations  
begin  
    concurrent statements  
end architecture;
```

0.2.4 arrays

```
type 'type_name' is array (range) of 'element_type';
```

Beispiele:

```
type speicher is array (0 to 1023) of integer range 0 to 255;  
signal sram : speicher;  
  
type mem8 is array (natural range <>) of std_logic_vector(7 downto 0);  
signal sram8_1024 : mem8(1023 downto 0);  
signal sram8_8 : mem8(7 downto 0);
```

Zugriff/Initialisierung:

```
-- beide gleichwertig, Elemente 1 und 2 werden mit "5" beglückt  
sram8_8 <= ("0", "0", "0", "0", "0", "5", "5", "0" );
```

```
sram8_8 <= ( 2 | 1 => "5", others => "0");

-- oder einzelnes Element mit 0xB füllen
sram8_8(2) <= x"B"
```

0.2.5 assert

```
assert 'bedingung' report 'string' severity 'severity_level';
```

Überwache, dass *bedingung* erfüllt ist, wenn NICHT report *string* mit severity *severity_level*

severity_level ist "error" (default), "note", "warning" oder "failure"

Beispiel:

```
assert (a > c) report "a muss grösser c sein" severity note;

assert (true) report "diese Meldung wird nie ausgegeben" severity note;
assert (false) report "diese Meldung wird immer ausgegeben" severity note;
```

0.2.6 attributes

T repräsentiert ein beliebiger Typ, A repräsentiert ein Array, S repräsentiert ein beliebiges Signal und E repräsentiert eine Entity.

Liste von vordefinierten Attributen	
Attribut	Beschreibung
T'BASE	Basistyp von T
T'LEFT	Wert am weitesten links in T. (Grösster wenn downto)
T'RIGHT	Wert am weitesten rechts in T. (Kleinster wenn downto)
T'HIGH	Grösster Wert in T.
T'LOW	Kleinster Wert in T.
T'ASCENDING	Boolean, TRUE wenn range definiert mit "to".
T'IMAGE(X)	String der den Wert X repräsentiert.
T'VALUE(X)	Wert vom Typ T, konvertiert vom String X.
T'POS(X)	Integer Position von X im diskreten Typ T.
T'VAL(X)	Wert vom diskreten Typ an integer Position X.
T'SUCC(X)	Wert vom diskreten Typ der auf X folgt
T'PRED(X)	Wert vom diskreten Typ der vor X liegt
T'LEFTOF(X)	Wert vom diskreten Typ links von X
T'RIGHTOF(X)) Wert vom diskreten Typ rechts von X
A'LEFT	Eintrag ganz links in A
A'LEFT(N)	Eintrag ganz links in Dimension N von A
A'RIGHT	Eintrag ganz rechts in A
A'RIGHT(N)	Eintrag ganz links in Dimension N von A

Liste von vordefinierten Attributen	
Attribut	Beschreibung
A'HIGH	Höchster Eintrag in A
A'HIGH(N)	Höchster Eintrag in Dimension N von A
A'LOW	Tiefster Eintrag in A
A'LOW(N)	Tiefster Eintrag in Dimension N von A
A'RANGE	Range von A'LEFT to A'RIGHT oder A'LEFT downto A'RIGHT
A'RANGE(N)	Range von Dimension N in A
A'REVERSE_RANGE	Range in A to und downto umgekehrt
A'REVERSE_RANGE(N)	REVERSE_RANGE von Dimension N in A
A'LENGTH	Integer Wert der Anzahl Elemente in A
A'LENGTH(N)	Anzahl Werte in Dimension N von A
A'ASCENDING	Boolean, TRUE wenn range definiert mit "to"
A'ASCENDING(N)	Boolean, TRUE wenn Dimension N in A definiert mit "to"
S'DELAYED(t)	Signalwert zur Zeit NOW -t
S'STABLE	TRUE, wenn kein Event in S
S'STABLE(t)	TRUE, wenn kein Event in S für Zeit t
S'QUIET	TRUE, wenn kein Event in diesem Simulations Zyklus
S'QUIET(t)	TRUE, wenn kein Event in S für Zeit t
S'TRANSACTION	Bit Signal, invertiert immer wenn Signal S ändert
S'EVENT	TRUE, wenn Signal S Event in diesem Simulationszyklus hatte
S'ACTIVE	TRUE, wenn Signal S aktiv in diesem Simulationszyklus
S'LAST_EVENT	Zeit seit letztem Event auf Signal S
S'LAST_ACTIVE	Zeit seit Signal S zuletzt aktiv
S'LAST_VALUE	Vorhergehender Wert von Signal S
S'DRIVING	
S'DRIVING_VALUE	
E'SIMPLE_NAME	String mit Name der Entity E
E'INSTANCE_NAME	String mit Designs Hierarchie inkl. Entity E
E'PATH_NAME	String zu Design Wurzel von E

Beispiele:

signal'LEFT	7	bei std_logic_vector(7 downto 0);
	0	bei std_logic_vector(0 to 7);
signal'RIGHT	0	bei std_logic_vector(7 downto 0);
	7	bei std_logic_vector(0 to 7);
signal'HIGH	7	bei std_logic_vector(7 downto 0);
	7	bei std_logic_vector(0 to 7);
signal'LOW	0	bei std_logic_vector(7 downto 0);
	0	bei std_logic_vector(0 to 7);
signal'RANGE	7 downto 0	bei std_logic_vector(7 downto 0);
	0 to 7	bei std_logic_vector(0 to 7);
signal'REVERSE_RANGE	0 to 7	bei std_logic_vector(7 downto 0);
	7 downto 0	bei std_logic_vector(0 to 7);
signal'LENGTH	8	bei std_logic_vector(7 downto 0);

```

            8          bei std_logic_vector(0 to 7);
signal'EVENT      if (clk'event and clk='1') then

```

0.2.7 block statements

```

''block_name'': block
  declarations
begin
  concurrent statements
end block;

```

0.2.8 case

```

case ''expression'' is
  when ''fall_1'' => ''sequential statement''
  when ''fall_2'' => ''sequential statement''
  when others    => ''sequential statement''
end case;

```

Beispiel1:

```

case wert is
  when 0    => w <= '1';
  when 1    => w <= '0';
  when 2 | 3 => w <= a;
  when 4 to 7 => w <= b;
  when others => w <= 'X';
end case;

```

Beispiel2:

```

wert <= '0';
case din is
  when "00" => wert <= '1';
  when others => null;           -- "when others" soll immer vorhanden sein
end case;                       -- durch default-Zuweisung is "null"-statement
möglich!

```

Beispiel3:

```

type state_type is ( IDLE, DO_SOMETHING );
...
case state is
  when IDLE      => tx_line <= '0';
  when DO_SOMETHING => tx_line <= '1';
  when others =>
    assert false report "case defaulted!" severity failure;
end case;

```

0.2.9 component declaration

Deklaration zur Festlegung des Namens und der Schnittstelle einer Komponente, die einer Entitydeklaration und Architecture zugeordnet sein muss.

```
component 'component_name'  
  generic ('generic_liste');  
  port ('port_liste');  
end component;
```

0.2.10 component instantiation

```
label: 'component_name'  
generic map ( 'generic1' => ' generic1_entity' )  
port map (  
  'component_port1' => 'entity_port1',  
  'component_port2' => 'entity_port2',  
  ...  
  'component_portx' => 'entity_portx'  
);
```

0.2.11 constant

```
constant 'constant_name' : type := value;
```

Beispiel:

```
constant festwert : std_logic_vector(7 downto 0) := "10101100";  
constant zeitwert : time := 50 ns;  
  
type rdatum is array (0 to 3) of bit_vector(7 downto 0);  
constant rom : rdatum :=  
  ("00000001",  
   "00000010",  
   "00000011",  
   "00000100");
```

0.2.12 entity

Eine Struktureinheit in einem VHDL- Entwurfssystem. Beschreibt die Schnittstellen eines VHDL-Funktionsblocks nach außen. Mit Hilfe von Port-Anweisungen erfolgt die Deklaration der Anschlüsse innerhalb der Entity. Zu jeder Entity gehört eine Architecture.

```
entity 'entity_name' is  
  generic (generic_list);  
  port (port_list);  
end 'entity_name';
```

0.2.13 exit-Anweisung

mit der exit-Anweisung wird die "innerste" Schleife verlassen und mit der Anweisung, die direkt auf die Schleife folgt, fortgefahren.

Beispiel: Bestimmen der Anzahl der führenden Nullen

```
for i in signal'range loop
  exit when signal(i)='1';
  null_v := null_v + 1;
end loop;
```

0.2.14 file declaration

Beispiel:

```
library ieee;
  use ieee.std_logic_1164.all;
  use ieee.std_logic_textio.all;
library std;
  use std.textio.all;
```

```
architecture sim of dut is
  file my_file : text open write_mode is "my_file.dat";
begin
  proc: process(clk)
    variable outline : line;
    variable counter : integer := 0;
  begin
    if rising_edge(clk) then
      write(outline, string'("Takt: "));
      write(outline, counter);
      writeline(my_file, outline);
      counter := counter + 1;
    end if;
  end process proc;
end sim;
```

0.2.15 for loop

```
'eventuell_label': for 'parameter' in 'range' loop
  sequential statements
end loop 'eventuell_label';
```

- Parameter muss nicht deklariert werden.
- Parameter darf im loop nicht verändert werden

Muss loop synthetisierbar sein:

- range ist statisch
- keine wait statments im loop

Beispiel:

```
for i in 0 to 7 loop
  w(i) <= a(i) and b;  -- 8bit bus "a" mit Einzelsignal "b" verunden
end loop;
```

Beispiel 2:

```
for i in 1 to 10 loop
  if (REPEAT = '1') then
    i := i-1;  -- Error
  end if;
end loop;
```

0.2.16 functions

Eine der beiden Möglichkeiten in VHDL, Code mittels eines einfachen Aufrufmechanismus wiederverwertbar zu machen. Functions werden gewöhnlich mit ihrem Namen und einer in Klammern stehenden Liste der Eingangsparameter aufgerufen und können nur ein Ausgangsargument liefern- vgl. auch Procedure.

```
function 'funktion_name' (parameter_list) return 'type' is
  declarations
begin
  sequential statements
end funktion_name;
```

Beispiel:

```
function parity_generator (din : std_ulogic_vector)
  return std_ulogic is
  variable t : std_ulogic := '0';  -- variable mit default Zuweisung
begin
  for i in din'range loop          -- ganze Busbreite
    t := t xor din(i);
  end loop;
  return t;
end parity_generator;
```

Aufruf der Funktion als "concurrent" oder "sequential statement":

```
sig_pary <= parity_generator(data_bus);
```

Achtung: keine "signal assignments" oder "wait"

0.2.17 generate

```
'label': for 'parameters' in 'range' generate
  concurrent statements
end generate 'label';
```

oder

```

'label': if 'condition' generate
  concurrent statements
end generate 'label';

```

Beispiel:

```

architecture gen of test is

  component volladdierer
    port (x,y,ci : in bit;
          s,co  : out bit);
  end component;

  component halbaddierer
    port (x,y : in bit;
          s,co : out bit;
  end component;

  signal carry : bit_vector(0 to 7);

begin

  gen_addierer: for i in 0 to 7 generate

    niedrigstes_bit: if i=0 generate
      w0: entity halbaddierer port map
        (x(i),y(i),s(i),carry(i));
    end generate niedrigstes_bit;

    hoeheres_bit: if i>0 generate
      w1: entity volladdierer port map
        (x(i),y(i),carry(i-1),s(i),carry(i));
    end generate hoeheres_bit;

  end generate gen_addierer;

  co <= carry(7);

end gen;

```

Beispiel2:

```

library ieee;
use ieee.std_logic_1164.all;
use ieee.numeric_std.all;
...
dft_mem_in      : in  std_logic_vector(const1 downto 0);
dft_mem_out     : out std_logic_vector(const2 downto 0);
pwr_mem_ctrl_in : in  std_logic_vector(3 downto 0)
...
gen0: for i in dft_mem_out'range generate
  signal tmp : std_logic_vector(dft_mem_out'range);
begin
  tmp <= std_logic_vector(resize(unsigned(dft_mem_in), dft_mem_out'length));
  dft_mem_out(i) <= tmp(i) xor (pwr_mem_ctrl_in(0) xor pwr_mem_ctrl_in(1) xor
pwr_mem_ctrl_in(2) xor pwr_mem_ctrl_in(3));
end generate gen0;

```

0.2.18 generic

```
entity 'entity_name' is
  generic (generic_list);
  port    (port_list);
end 'entity_name';
```

Beispiel: (wichtig für skalierbare Blöcke!!)

```
entity test is
  generic (n : integer := 15); -- hierbei ist 15 der Default-Wert, falls kein
Generic
  -- bei der Initialisierung angegeben wird
  port    (a : in std_ulogic_vector(n-1 downto 0));
end test;
```

0.2.19 if

```
if    condition_a then
  sequential statements
elsif condition_b then
  sequential statements
else
  sequential statements
end if;
```

Beispiel:

```
if nreset='0' then
  count <= 0;
elsif clk'event and clk='1' then
  if count=9 then
    count <= 0;
  else
    count <= count+1;
  end if;
end if;
```

0.2.20 library

0.2.21 names

0.2.22 next-Anweisung

Die next-Anweisung beendet den aktuellen Schleifendurchlauf vorzeitig; das bedeutet, dass die Anweisungen bis zur end-loop-Anweisung übersprungen werden und mit dem nächsten Schleifendurchlauf fortgefahren wird.

Beispiel: Bestimmen der Anzahl der Nullen in einem Vektor

```

for i in signal'range loop
  next when signal(i)='1';
  null_v := null_v + 1;
end loop;

```

0.2.23 notations

```

hex_var := 16#8001#;

binary_s <= b"000_111_010";
octal_s   <= o"207";
hex_s    <= x"01_FB";

```

0.2.24 null statement

Falls durch die Syntax ein Statement erforderlich ist, kann das "Null"-Statement verwendet werden, um anzuzeigen, dass nichts zu tun ist. Vgl. hierzu beispielsweise 'when others => null;' einer case-Anweisung.

```

proc: process(clk)
begin
  if rising_edge(clk) then
    case select is
      when "00" => reg <= '1';
      when "11" => reg <= '0';
      when others => null;
    end case;
  end if;
end process proc;

```

0.2.25 operators

Logische Operatoren für Typen: bit,boolean,bit_vector,std_logic,std_logic_vector

```

and      -- und
or       -- oder
nand     -- nicht und
nor      -- nicht oder
xor      -- exclusive oder
xnor     -- exclusives nicht oder

```

Vergleichs Operatoren Ergebnis: boolean

```

=        -- Gleichheit
/=       -- Ungleichheit

<        -- kleiner
>        -- groesser
<=       -- kleiner gleich (Achtung bei Type int: Speichern)

```

>= -- grösser gleich

Arithmetische Operatoren für Typen: integer,real

```
a <= a + 7;      -- Addition
r1 <= r2 - 3.1415 -- Subtraktion (real)
m <= x * y      -- Multiplikation
d <= m / 2      -- Division
```

VHDL93:

```
sll -- shift left  logical
srl -- shift right logical
sla -- shift left  arith.
sra -- shift right arith.
rol -- rotate left
ror -- rotate right
```

0.2.26 package

```
package 'package_name' is
  declarations
end package;
```

Beispiele:

```
package demo is
  constant nullwert : bit_vector := "00000000";
  function foo ( v : std_ulogic ) return std_ulogic;
  component adder      -- Dessen Implementierung ist vielleicht in
  irgendeiner Library vorcompiliert,
  port(x,y,ci : in bit; -- da ein Component niemals in der package body
  definiert werden kann.
        s,co  : out bit);
  end component;
end demo;
```

```
package body demo is
  function foo ( v : std_ulogic ) return std_ulogic is
  begin
    return v;
  end function;
end package body;
```

Package-Aufruf lautet dann:

```
use work.demo.all;

entity xx is
  port
    ( wert : out bit_vector(7 downto 0));
end xx;

architecture behv of xx is
  begin
    wert <= nullwert;
```

```
end behv;
```

0.2.27 procedures

-- Parameter können constant, variable oder Signale sein. Auf Signale kann gelesen (in) oder geschrieben (out) werden.

```
procedure 'procedure_name' (paramter_list) is
  declarations
begin
  sequential statements
end 'procedure_name';
```

Beispiel:

```
procedure parity_generator
  (signal din : in std_ulogic_vector;
   signal par : out std_ulogic) is
  variable t : std_ulogic := '0';
begin
  for i in 0 to din'range loop
    t := t xor din(i);
  end loop;
  par <= t;
end parity_generator;
```

Procedures in Packages:

```
package my_package is

  procedure parity_generator
    (signal din : in std_ulogic_vector;
     signal par : out std_ulogic);

end my_package;

package body my_package is

  procedure parity_generator
    (signal din : in std_ulogic_vector;
     signal par : out std_ulogic) is

    variable t : std_ulogic := '0';
  begin
    for i in 0 to din'range loop
      t := t xor din(i);
    end loop;
    par <= t;
  end parity_generator;

end my_package;
```

Aufruf:


```
signal value_u   : unsigned(3 downto 0);
signal value_slv : std_logic_vector(3 downto 0);
--
value_i   <= 14;
value_u   <= to_unsigned(value_i, value_u'length); -- Typumwandlung
value_slv <= std_logic_vector(value_u);           -- cast (Typen sind eng
verwandt)
```

0.2.33 type declaration

```
type <memory_type> is array(0 to 9) of std_logic_vector(7 downto 0);
type <fsm_type> is (idle, run, ready);
type <hour_range_type> is range 1 to 12;
subtype <vector_type> is std_logic_vector(5 downto 0);
```

0.2.34 use

Verwendung von Bibliotheken, oder Teilen daraus.

Beispiel:

```
library ieee;
use ieee.std_logic_1164.all;
use ieee.numeric_std.all;
```

0.2.35 variable declaration

```
variable 'Variablenname' : 'Typ' := 'Initialisierungswert'
```

Beispiel:

```
variable MeinBool : boolean := false;
```

0.2.36 variable assignment

```
'Variablenname' := 'Wert';
```

0.2.37 wait

- wait until *condition*
- wait on *signal list*
- wait for *time* -- nicht synthetisierbar

- wait;

Beispiel:

```
wait until din="0010";
```

oder:

```
stimulus: process
begin
    loop
        clk <= '0';
        wait for 50 ns;
        clk <= '1';
        wait for 50 ns;
    end loop;
end process;
```

0.2.38 when

```
signal_name <= expression when condition else
    expression when condition else
    expression;
```

Beispiel (Tristate Port):

```
data <= data_out when data_enable = '1' else 'Z';
```

0.2.39 while

```
while condition loop
    sequential statements
end loop;
```

Beispiel (Register rechts schieben):

```
if clk'event and clk='1' then
    i:=0;
    while i<7 loop
        buswert(i) <= buswert(i+1);
        i:=i+1;
    end loop;
    buswert(7) <= din;
end if;
```

0.2.40 with select

```
with select _signal select
  dst_signal <= src_signal when select_value1,
               src_signal when select_value2,
               src_signal when others;
```

Selektive Signalzuweisung außerhalb eines Prozesses, als Alternative zu case:

```
signal sel_s          : std_logic_vector(1 downto 0);
signal monitor_s,one_s,two_s : std_logic_vector(3 downto 0);
```

```
with sel_s select
  monitor_s <= one_s when "00",
             two_s when "11",
             "0000" when others;
```

0.3 Operator precedence

VHDL operators in order of precedence (Highest first)

Note: The concatenate operator & has a lower order of precedence than some arithmetic operators +/-

```
**
abs
not
*
/
mod
rem
+
-
+
-
&
sll
srl
sla
sra
rol
ror
=
/=
<
<=
>
>=
and
or
nand
nor
xor
xnor
```

0.4 Beispielprogramme

0.4.1 Zustandsmaschinen

Die im Folgenden beschriebene Codierung einer Zustandsmaschine erhebt nicht den Anspruch die eleganteste Lösung darzustellen, sie soll vielmehr einen Eindruck der Sprache und der grundsätzlichen Abläufe vermitteln.

```
architecture demo of zustandsmaschine is

    constant vectorbreite : integer := 3;

    type zustandsvector is std_logic_vector(vectorbreite downto 0);

    signal zustand : zustandsvector;

    -- "one hot" Codierung !!
    constant warte_zustand : zustandsvector := "0001";
    constant a_zustand : zustandsvector := "0010";
    constant b_zustand : zustandsvector := "0100";
    constant c_zustand : zustandsvector := "1000";

    signal gehe_zu_warte_zustand : std_logic;
    signal gehe_zu_a_zustand : std_logic;
    signal gehe_zu_b_zustand : std_logic;
    signal gehe_zu_c_zustand : std_logic;

    signal timer : std_logic_vector(3 downto 0);

begin
    -----
    zustaende : process(nres,clk)
    begin
        if (nres='0') then
            zustand <= warte_zustand;
        elsif (clk'event and clk='1') then
            if (gehe_zu_warte_zustand='1') then
                zustand <= warte_zustand;
            elsif (gehe_zu_a_zustand='1') then
                zustand <= a_zustand;
            elsif (gehe_zu_b_zustand='1') then
                zustand <= b_zustand;
            elsif (gehe_zu_c_zustand='1') then
                zustand <= c_zustand;
            else
                -- der "else" Pfad ist optional!
                zustand <= zustand; -- ohne diese Zeilen identische Funktion!!
            end if;
        end process zustaende;
    -----
    zustandsweitschaltung : process(ingang1,ingang2,zustand,timer)
    begin
        gehe_zu_warte_zustand <= '0'; -- default assignments
        gehe_zu_a_zustand <= '0';
        gehe_zu_b_zustand <= '0';
        gehe_zu_c_zustand <= '0';

        case zustand is
            when warte_zustand =>
                if (ingang1='1') then -- Priorisierung von
                    gehe_zu_a_zustand <= '1'; -- "ingang1" und
"ingang2"
                elsif (ingang2='1') then
                    gehe_zu_c_zustand <= '1';
                end if;
    end process zustandsweitschaltung;
end architecture demo;
```

```

    when a_zustand    =>
        if (timer="0000") then          -- Maschine bleibt im
"a_a_zustand"
            gehe_zu_b_zustand <= '1';  -- bis Timer abgelaufen
            end if;
    when b_zustand    =>
        gehe_zu_warte_zustand <= '1';
    when c_zustand    =>
        gehe_zu_a_zustand <= '1';
    when others       =>                -- diese Zeile sichert
Vollständigkeit
        gehe_zu_warte_zustand <= '1';  -- der Auscodierung!!!
    end case;
end process zustandsweitzerschaltung;
-----
wartezeit : process (nres,clk)
begin
    if (nres='0') then
        timer <= "1010";
    elsif (clk'event and clk='1') then
        if (zustand=a_zustand) then    -- Zahler steht auf "1010"
            timer <= timer-1;         -- nur im "a_zustand" läuft er rückwärts
        else
            timer <= "1010";
        end if;
    end if;
end process wartezeit;
-----
getakteter_ausgang : process (nres,clk)
begin
    if (nres='0') then
        ausgang1 <= '0';
    elsif (clk'event and clk='1') then
        if (zustand=b_zustand) then
            ausgang1 <= '1';          -- eine "clk"-Periode langer Pulse
        else                          -- Achtung: erscheint einen Takt nach
            "b_zustand"!!
            ausgang1 <= '0';
        end if;
    end if;
end process getakteter_ausgang;
-----
asynchroner_ausgang : process (gehe_zu_b_zustand)
begin
    ausgang2 <= gehe_zu_b_zustand;    -- eine "clk"-Periode langer Pulse (könnte
    spiken!!)
end process asynchroner_ausgang;
-----

```

0.5 Fehlervermeidung

0.5.1 Unbeabsichtigtes Erzeugen eines "Latch"

In der Synthese wird unbeabsichtigt ein "Latch" implementiert. Die Ursache ist meist, dass in einem kombinatorischen Prozess die Zuweisungen auf ein Signal nicht vollständig auscodiert wurden:

Beispiel:

```
if (a='1') and (b='0') then
  x <= '1';
elsif (a='0') and (b='1') then
  x <= '0';
end if;
```

Die Fälle a=1 und b=1 ebenso wie a=0 und b=0 wurden nicht definiert. Die Folge ist, dass die Synthese versucht in diesen Fällen den aktuellen Zustand beizubehalten. Dies erfolgt durch die Implementierung eines "Latch".

0.5.2 Vergessen von Signalen in der "sensitivity list" eines kombinatorischen Prozesses

Werden Signale in der Liste nicht absichtlich weggelassen, um ein bestimmtes Verhalten der Schaltung zu erzeugen, sondern aus Nachlässigkeit nicht hinzugefügt, wird als Folge das Verhalten der Simulation von dem der realen Gatterschaltung abweichen. Es gibt vhdl-Editoren, die die "sensitivity list" prüfen. Auch in der Synthese erscheinen meistens Warnmeldung.

en:Programmable Logic/VHDL¹ fr:Conception et VHDL²

0.5.3 Verwenden von Reset-Signalen in einem Design

Asynchrone Reset-Signale gehören bei VHDL-Designs unbedingt synchronisiert:

```
library ieee;
use ieee.std_logic_1164.all;

entity myEnt is
  port(
    rst_an : in std_logic;
    clk: in std_logic;
    rst: in std_logic;
    sigIn: in std_logic_vector(3 downto 0);
    sigOut: out std_logic_vector(3 downto 0));
end entity myEnt;

architecture myArch of myEnt is
```

¹ <http://en.wikibooks.org/wiki/Programmable%20Logic%2FVHDL>

² <http://fr.wikibooks.org/wiki/Conception%20et%20VHDL>

```
signal sync_rst_r : std_logic_vector(1 downto 0);
signal mySig: std_logic_vector(sigIn'range);

begin

process(clk)
begin
if (rising_edge(clk)) then
sync_rst_r <= sync_rst_r(0) & rst_an;
end if;
end process;

process(clk, sync_rst_r)
begin
if sync_rst_r(1) = '0' then
mySig <= (others => '0');
elsif (rising_edge(clk)) then
if (rst = '1') then
mySig <= (others => '0');
else
mySig <= sigIn;
end if;
end if;
end process;

sigOut <= mySig;

end architecture;
```

Für detaillierte Information zum Thema Reset im FPGAs siehe das Whitepaper von Xilinx "Get Smart About Reset: Think Local, Not Global" (englischsprachig).

0.5.4 Generieren von "Clock"-Signalen in einem Design

Die "Clock"-Signale werden mit Hilfe einer speziellen Verdrahtung auf dem FPGA verteilt. Ein "Clock"-Signal darf nie durch Logik erzeugt werden. Im Folgenden wird ein schlechtes Beispiel gezeigt, in welchem ein neues Clock Signal erzeugt wird:

```
library ieee;
use ieee.std_logic_1164.all;

entity clock_divider_ent is
port (
clk      : in    std_logic;
clkDiv   : out   std_logic);
end clock_divider_ent;

architecture synth of clock_divider_ent is

signal counter: integer range 1023 downto 0;
signal clkDivInt: std_logic := '0';

begin

process(clk)
begin
if (rising_edge(clk)) then
if (counter = 0) then
counter <= 1023;
clkDivInt <= not clkDivInt;
else
```

```

        counter <= counter - 1;
    end if;
end if;
end process;

end synth;

```

In diesem Beispiel wird eine bessere Implementierung für den oben gezeigten Block dargestellt. Das Signal `clkDivInt` wird hier zu einem kurzen Puls: jedes Mal wenn der Zähler den Wert null erreicht, bleibt dieser Puls hoch nur während eines einzigen Taktzyklus vom `clk`.

```

library ieee;
use ieee.std_logic_1164.all;

architecture synth of clock_divider_ent is

    signal counter:    integer range 2047 downto 0;
    signal clkDivInt:  std_logic := '0';

begin

    process(clk)
    begin
        if (rising_edge(clk)) then
            counter <= counter - 1;
            clkDivInt <= '0';
            if (counter = 0) then
                counter <= 2047;
                clkDivInt <= '1';
            end if;
        end if;
    end process;

end synth;

...
process(clk)
begin
    if(rising_edge(clk)) then
        if (clkDiv = '1') then
            ...
        end if;
    end if;
end process;
...

```

0.5.5 Vergleich von Zahlen

Nach Möglichkeit sollte immer auf `"=`" und nicht auf `"<`", `">`", `"<="` oder `">="` verglichen werden, da diese aufwendiger in Hardware zu implementieren sind.

0.5.6 `std_logic_arith` vs `numeric_std`

`std_logic_arith` wurde nicht standardisiert und ist nicht überall gleich implementiert. Stattdessen, nur `numeric_std` verwenden:

```
library ieee;
use ieee.std_logic_arith.all; -- Vermeiden, nicht standardisiert.
use ieee.numeric_std.all;    -- Die beiden zusammen gibt nur Ärger.
```

Kategorie: Buch³

³ <http://de.wikibooks.org/wiki/Kategorie%3ABuch>

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