

FPGA Carry Chain Adder (1A)

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FPGA Carry Chain Cell



$$s_i = (a_i \oplus b_i) \oplus c_i = p_i \oplus c_i$$

$$c_{i+1} = (a_i \cdot b_i) + (a_i \oplus b_i) c_i = \bar{p}_i \cdot g_i + p_i \cdot c_i = \bar{p}_i \cdot a_i + p_i \cdot c_i = \bar{p}_i \cdot b_i + p_i \cdot c_i$$

when $\bar{p}_i = 1$, then $a_i = b_i$

when $g_i = 1$, then $a_i = b_i = 1$

$p(i)$	0	1
0	0	1
1	1	0

$g(i)$	0	1
0	0	0
1	0	1

FPGA Carry Chain Cell



Synthesis of Arithmetic Circuits: FPGA, ASIC and Embedded Systems, J-P Deschamps et al

FPGA Carry Chain

FPGAs generally contain dedicated computation resources for generating fast adders

The Virtex family programmable arrays include logic gates (**XOR**) and **multiplexers** that along with the general purpose **lookup tables** allow one to build effective carry-chain adders

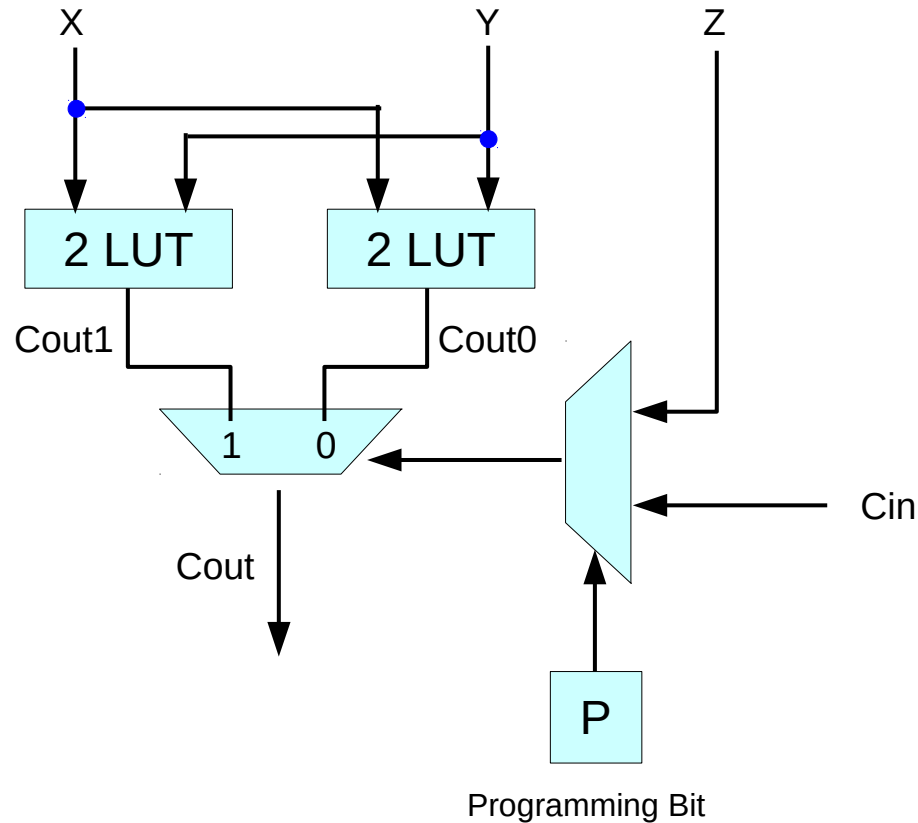
The carry chain is made up of multiplexers belonging to adjacent configurable blocks

the lookup table is used for implementing the exclusive or function

$$p(i) = x(i) \text{ xor } y(i)$$

https://en.wikipedia.org/wiki/Carry-lookahead_adder

FPGA Carry Chain Cell



Cout1, Cout2 : functions of X, Y, Cin

Cout1 = X+Y when Cin=1

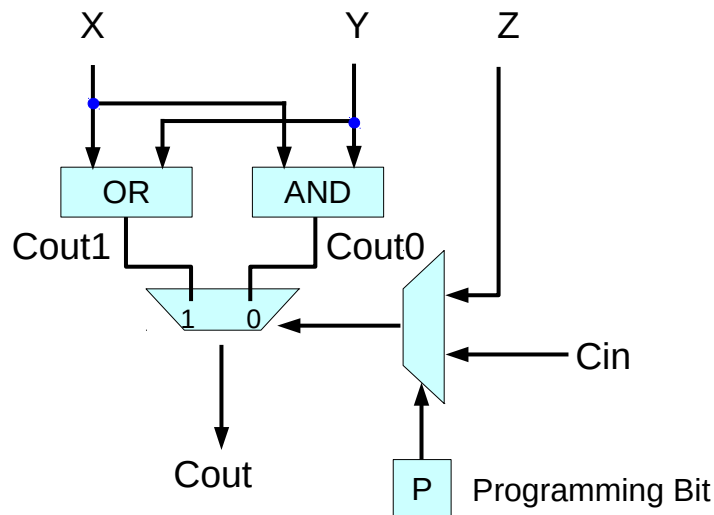
Cout0 = X Y when Cin=0

Cout = (X + Y) Cin + X Y $\overline{\text{Cin}}$

Cout = P' Cin + G $\overline{\text{Cin}}$... P' = relaxed P

Cout1	Cout0	Cout	Name
0	0	0	Kill
0	1	$\overline{\text{Cin}}$	Inverse Propagate
1	0	Cin	Propagate
1	1	1	Generate

FPGA Carry Chain Cell



X	Y	Cin	$\overline{\text{Cin}}$	$\overline{X} \overline{Y}$
		Cout1	Cout0	
0	0	0	0	$\overline{X} \overline{Y}$
0	1	1	0	$\overline{X} Y$
1	0	1	0	$X \overline{Y}$
1	1	1	1	$X Y$

Cout : functions of X, Y, Cin

$$\text{Cout}(X, Y, 1) = \text{Cout1} = X + Y$$

$$\text{Cout}(X, Y, 0) = \text{Cout0} = X Y$$

$$\text{Cout1} = X + Y \text{ when Cin}=1$$

$$\text{Cout0} = X Y \text{ when Cin}=0$$

$$\text{Cout1} = P' \text{Cin} \dots P' = \text{relaxed } P$$

$$\text{Cout0} = G \overline{\text{Cin}}$$

$$\text{If Cin, then Cout} = (\overline{X} Y + X \overline{Y} + X Y)$$

$$\text{If } \overline{\text{Cin}}, \text{ then Cout} = X Y$$

$$\text{Cin} (X + Y) + \overline{\text{Cin}} X Y$$

$$\text{Cin} (\overline{X} Y + X \overline{Y} + X Y) + \overline{\text{Cin}} X Y$$

$$\text{Cin} (\overline{X} Y + X \overline{Y}) + (\text{Cin} + \overline{\text{Cin}}) X Y$$

$$P \text{Cin} + G$$

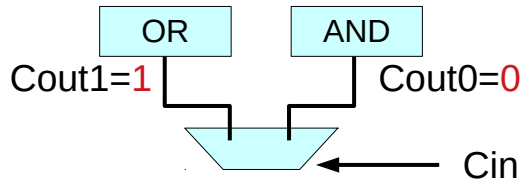
$$\text{Cin} (X + Y) + \overline{\text{Cin}} X Y$$

$$\text{Cin } P' + \overline{\text{Cin}} G$$

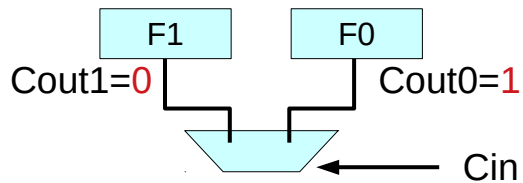
$$\dots P' : \text{relaxed } P$$

High Performance Carry Chains for FPGAs, S. Hauck, M. M. Hosler, T. W. Fry

FPGA Carry Chain Cell



Cout1=1 when Cin=1
 Cout0=0 when Cin=0
 Cout = Cin



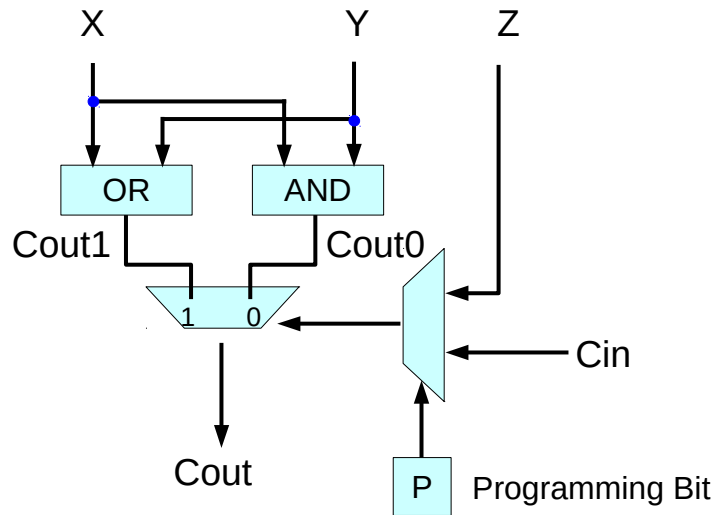
Cout1=0 when Cin=1
 Cout0=1 when Cin=0
 Cout = $\overline{\text{Cin}}$

Cout0	Cout1	Cout	Name
0	0	0	Kill
0	1	$\overline{\text{Cin}}$	Propagate
1	0	Cin	Inverse Propagate
1	1	1	Generate

Cout1	Cout0	Cout	Name
0	0	0	Kill
0	1	$\overline{\text{Cin}}$	Inverse Propagate
1	0	Cin	Propagate
1	1	1	Generate

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Carry Chain



Carry Out

X	Y	Cin	Cout
0	0	Cin	$\overline{\text{Cin}}$
0	1	Cin	$\overline{\text{Cin}}$
1	0	Cin	$\overline{\text{Cin}}$
1	1	Cin	Cin

X	Y	Cin	$\overline{\text{Cin}}$	$\overline{\text{X}} \overline{\text{Y}}$
		Cout1	Cout0	
0	0	0	0	$\overline{\text{X}} \overline{\text{Y}}$
0	1	1	0	$\overline{\text{X}} \text{Y}$
1	0	1	0	$\text{X} \overline{\text{Y}}$
1	1	1	1	$\text{X} \text{Y}$

Cout1	Cout0	Cout	Name
0	0	0	Kill
0	1	$\overline{\text{Cin}}$	Inverse Propagate
1	0	Cin	Propagate
1	1	1	Generate

Cout1=1 when Cin=1

Cout0=0 when Cin=0

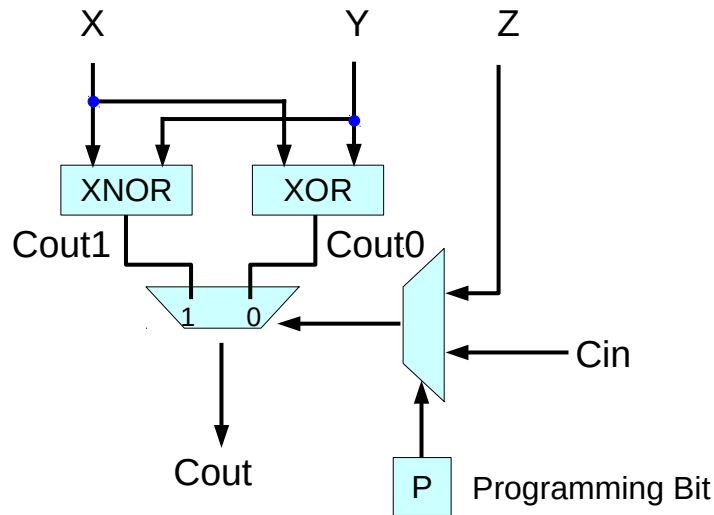
Cout = Cin propagate

Cout1=0 when Cin=1

Cout0=1 when Cin=0

Cout = $\overline{\text{Cin}}$ inverse propagate

Parity Checker



X	Y	Cin	$\overline{\text{Cin}}$	
		Cout1	Cout0	
0	0	1	0	$\overline{X} \overline{Y}$
0	1	0	1	$\overline{X} Y$
1	0	0	1	$X \overline{Y}$
1	1	1	0	$X Y$

Cout1	Cout0	Cout	Name
0	0	0	Kill
0	1	$\overline{\text{Cin}}$	Inverse Propagate
1	0	Cin	Propagate
1	1	1	Generate

Computing Parity

$X \oplus Y \oplus \text{Cin}$	
$0 \oplus 0 \oplus \text{Cin}$	$\overline{\text{Cin}}$
$0 \oplus 1 \oplus \text{Cin}$	$\overline{\text{Cin}}$
$1 \oplus 0 \oplus \text{Cin}$	$\overline{\text{Cin}}$
$1 \oplus 1 \oplus \text{Cin}$	Cin

Cout1=1 when Cin=1

Cout0=0 when Cin=0

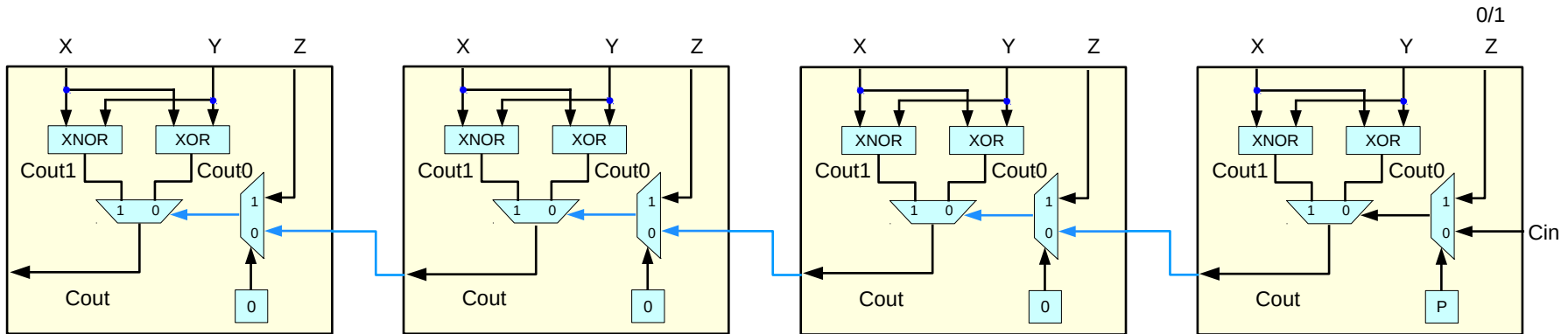
Cout = Cin propagate

Cout1=0 when Cin=1

Cout0=1 when Cin=0

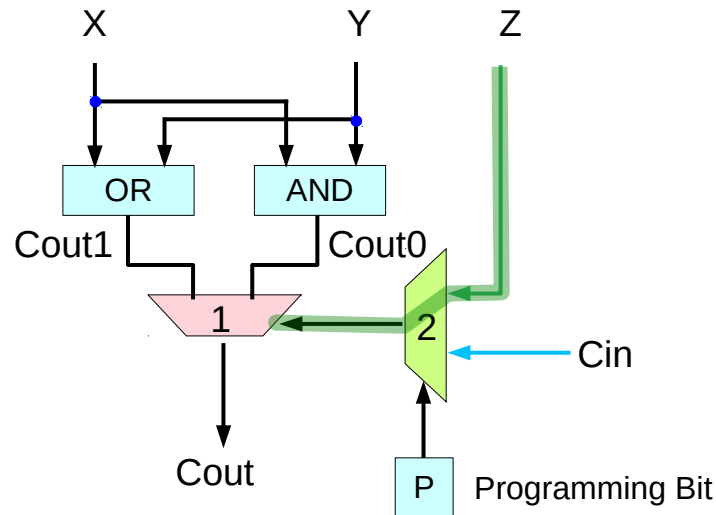
Cout = $\overline{\text{Cin}}$ inverse propagate

Ripple Carry Chain



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FPGA Carry Chain Cell



the logic cells - resources to compute a function
the exact location of logic cells depends on the user.
a user can start or end a carry computation
at any place in an fpga.

to start a carry chain, the **first cell** in the chain
must be programmed to **ignore** the **Cin** signal

program **mux2** in the cell to route input **Z**
to **mux1** instead of **Cin**

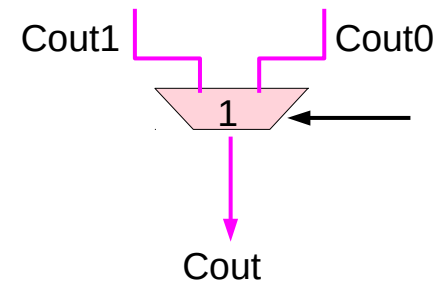
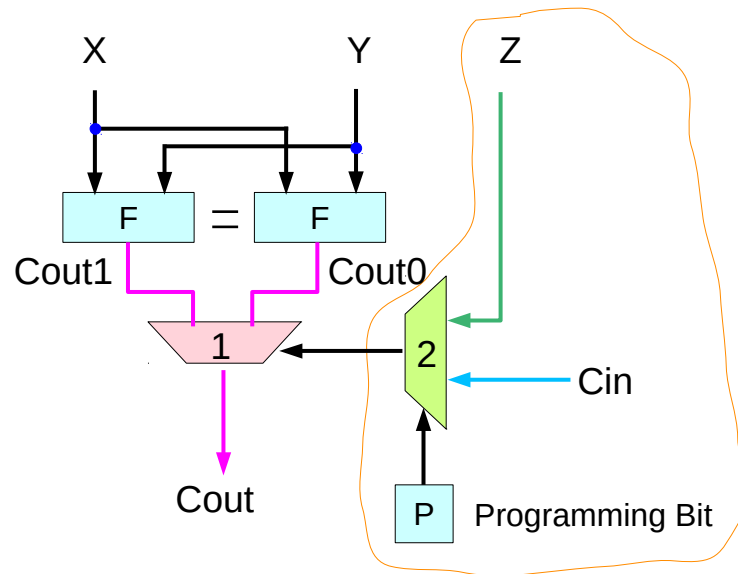
When it is desired to have a carry input
to the first cell of the chain
(implementing combined adder/subtractors)

But in many carry computations,
the first cell has only 2 inputs,
and forcing the carry chain
to wait for the arrival of an additional,
unnecessary input will only needlessly
slow down the circuit's computation.

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FPGA Carry Chain Cell

first cell



the first cell in the chain sometimes have only 2 inputs (X and Y) thus one of two 2-LUT's could compute this value

we can program both 2-LUTs with the same function, so that the output may have the proper value regardless of the Cin and Z inputs

can be routed to mux1 without changing the computation

however, this is only true if mux1 is implemented such that if the two inputs to the mux are the same, the output of the mux is identical to the inputs regardless of the state of the select line

Ripple Carry Chain

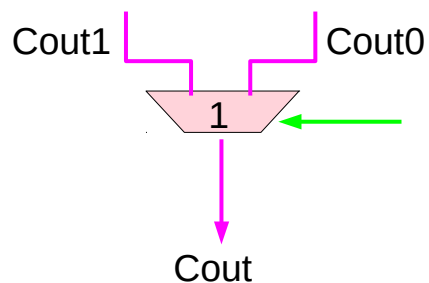
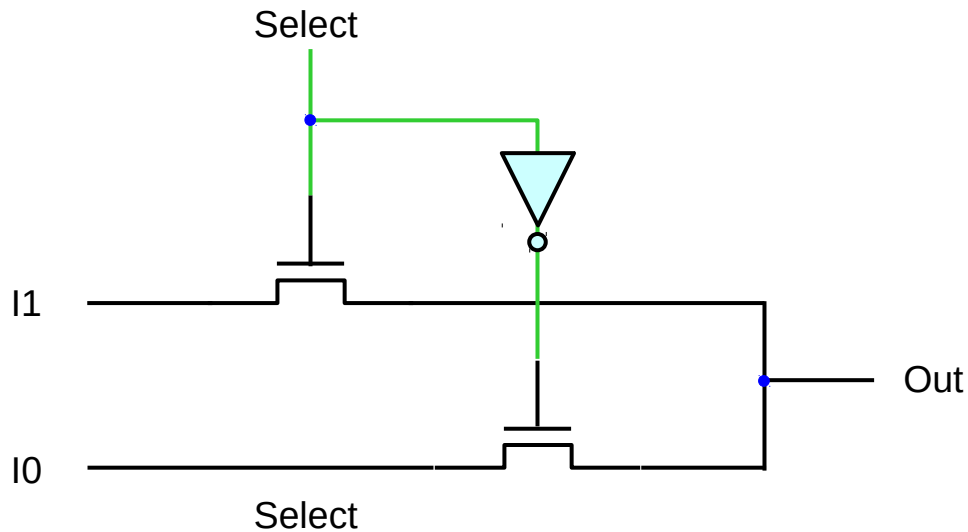


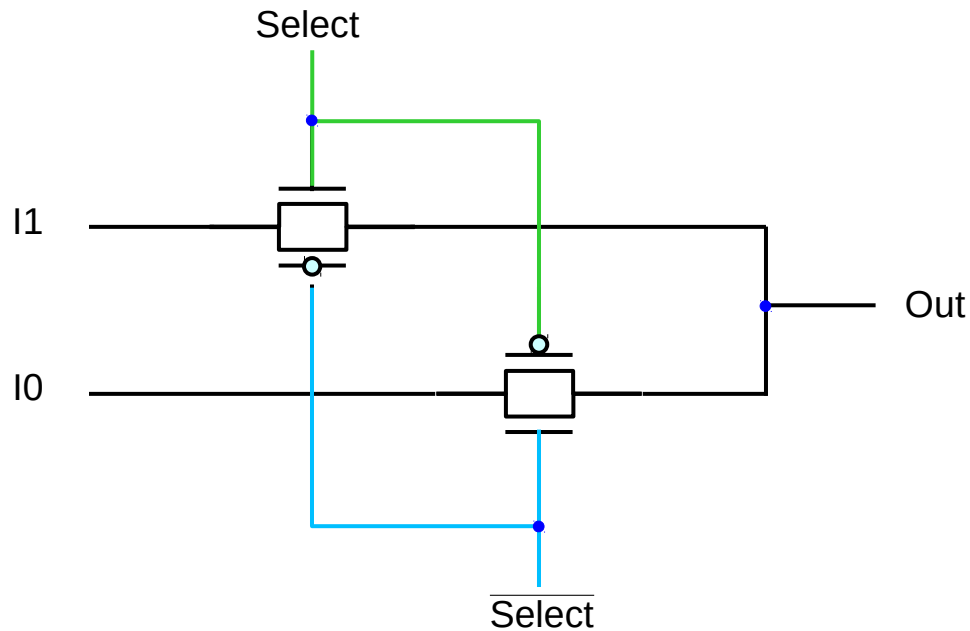
fig1b shows an implementation of a mux that does not obey this requirement

since the carry chain is part of an fpga, the input to this mux could be connected to some **unused logic** in another row which is generating **unknown values**.

if that unused logic had **multiple transitions** which caused the signal to change **quicker** than the gate could react, then it is possible that **the select signal** to this mux could be stuck midway between true and false (2.5V for 5V CMOS)

in this case, it will not be able to pass a true value from the input to the output and thus will not function properly for this application.

Ripple Carry Chain



however a mux built with both n-transistor and p-transistor pass gates will operate properly for this case

assume this mux implementation will be used

tristate driver based muxes could be used, which restore signal drive and cut series RC chains

Unit Gate Delay Model

All simple gate of two or three inputs that are directly implementable in **one logic level** in CMOS are considered to have a **delay of one**.

All other gate must be implemented by such gates, and have the delay of the underlying circuit.

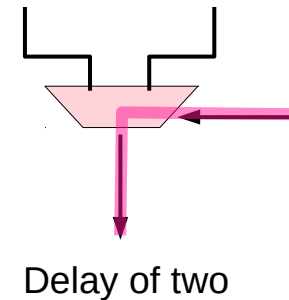
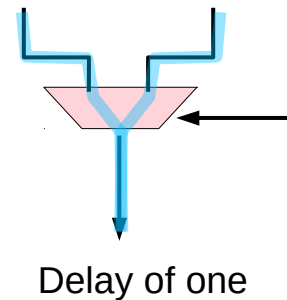
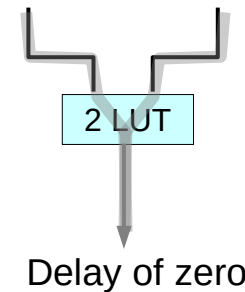
Delay of one

- inverters and
- 2 to 3 input NAND
- 2 to 3 input NOR gates

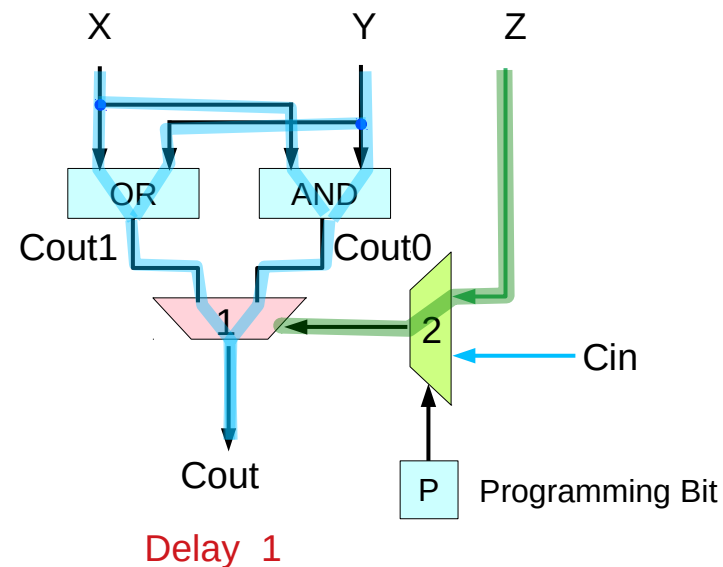
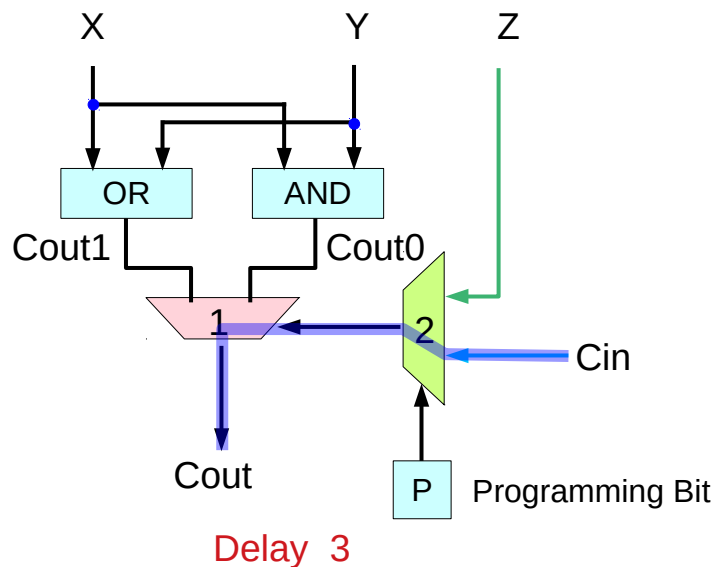
A **2:1 mux** has a **delay of one** from the I0 or I1 inputs to the output, But has a **delay of two** from the select input to the output due to the Inverter delay

Delay of zero (constant delay)

- the delay of the 2-LUTs,
- any routing leading to them,



FPGA Carry Chain Cell



Significantly slower
two muxes on the carry chain in each cell

Delay 1 for first cell
Delay 3 for each additional cell in the carry chain
1 delay for mux2 and
2 delays for mux1

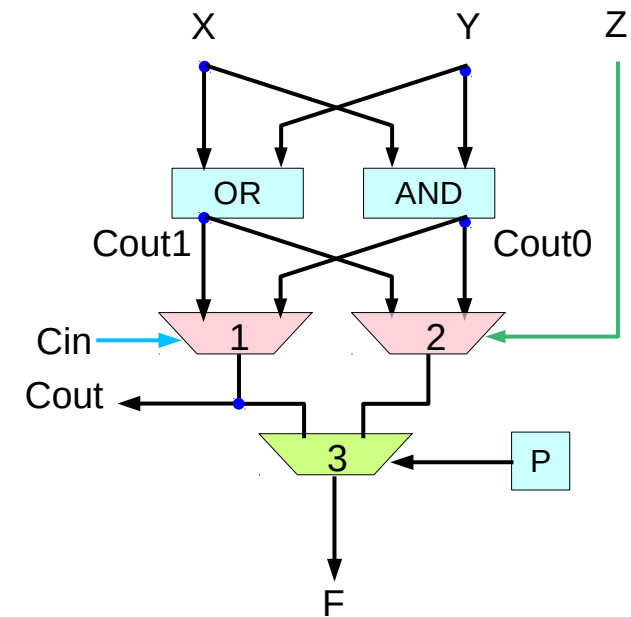
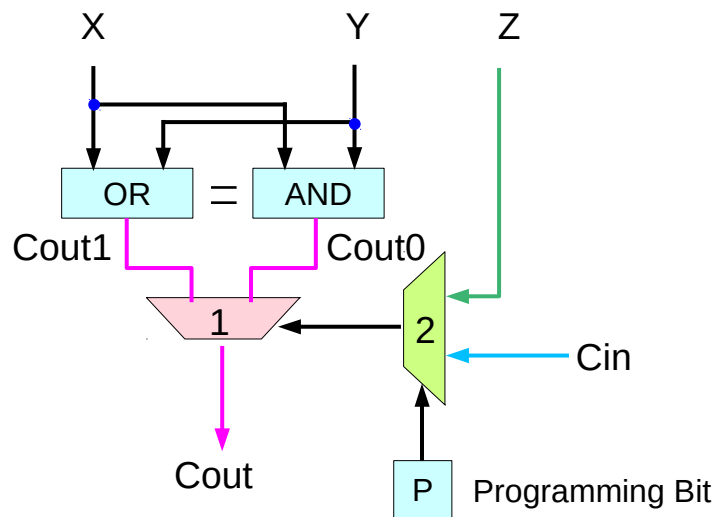
Overall $2n-2$ for an n -cell carry chain

The critical path comes from the 2-LUTs
and not input Z
since the delay through the 2-LUTs
will be larger than through mux 2 in the first cell

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FPGA Carry Chain Cell

- to reduce the delay of the ripple carry chain
 - remove mux2 from the carry path.
 - no need to choose between Cin and Z for the select line to the output mux1
 - two separate muxes, mux1 and mux2, controlled by Cin and Z respectively.
 - the circuit chooses between these outputs with mux3.

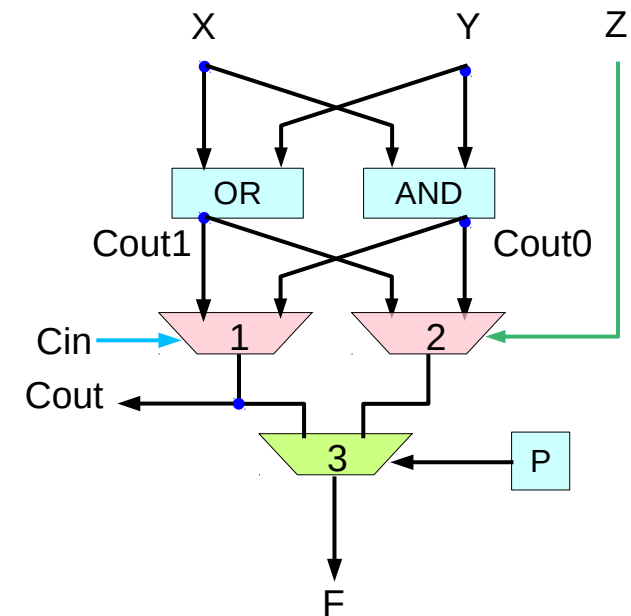


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FPGA Carry Chain Cell

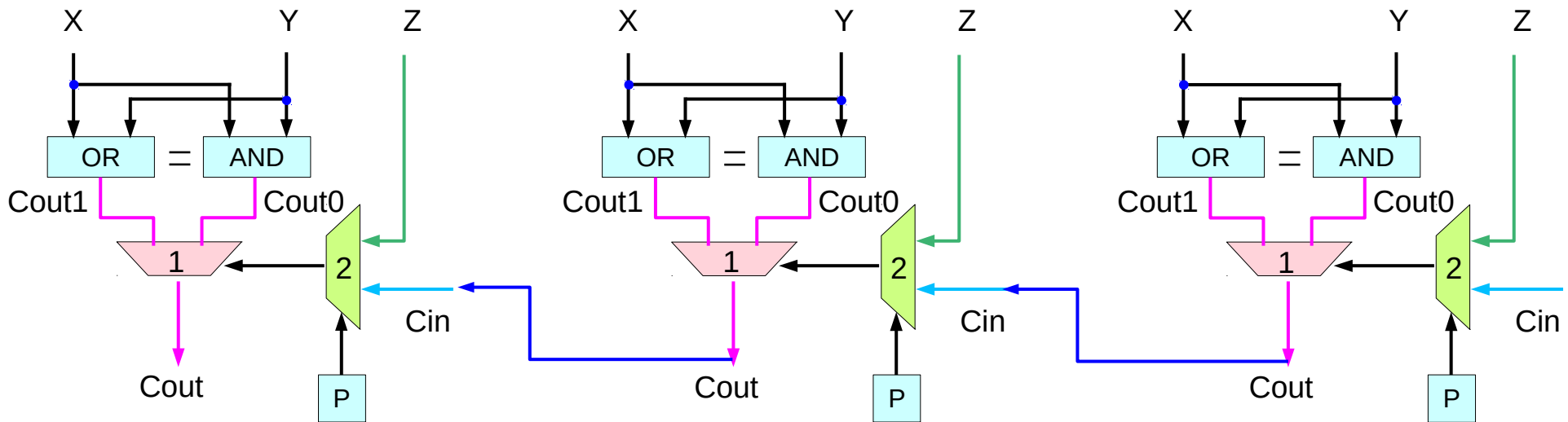
although this design is 1 gate delay slower than that of fig 2a, it provides the ability to have a carry input to the first cell in a carry chain, something that is important in many computations.

Also, for carry computations that do not need this feature, the first cell in a carry chain built from fig 2b can be configured to bypass mux1, reducing the overall delay to $2n$, which is identical to that of fig2a.



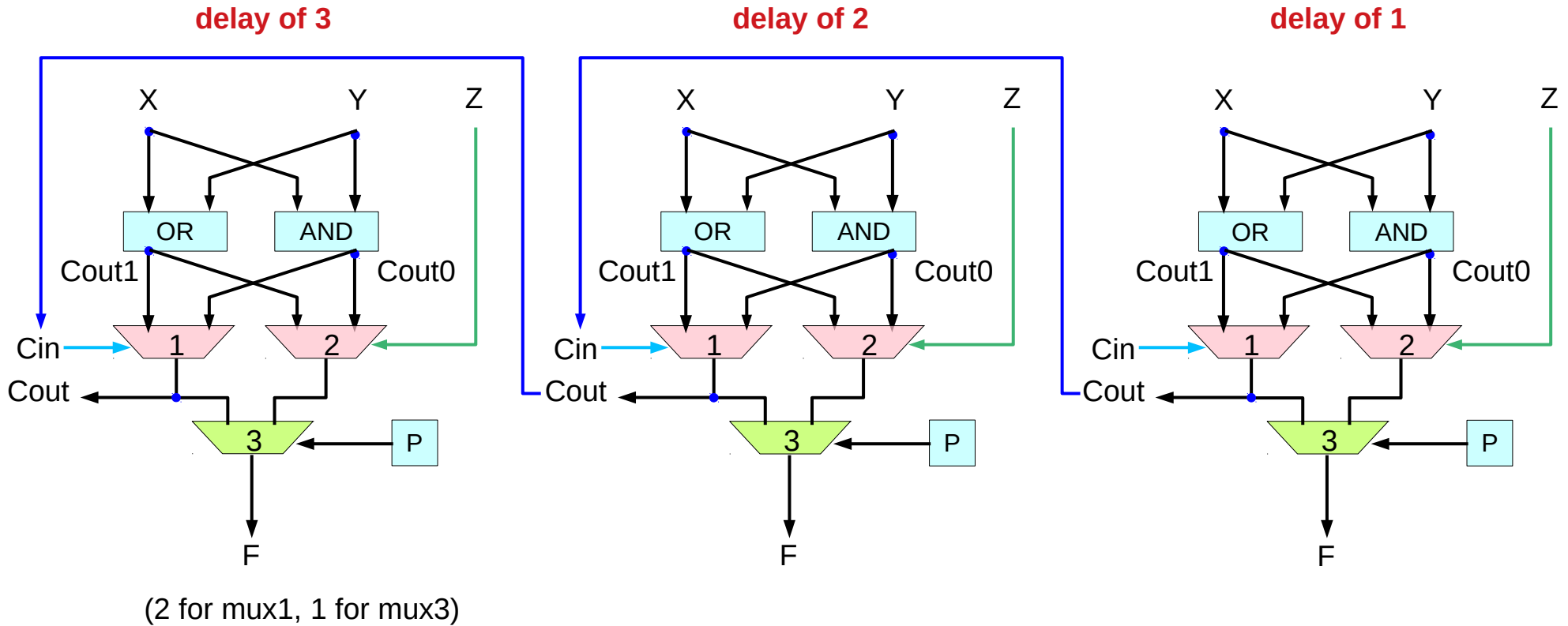
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FPGA Carry Chain Cell



on the other hand, in order to implement a n-bit carry chain with a carry input, the design of fig 2a requires an additional cell at the beginning of the chain to bring in this input, resulting in a delay of $2(n+1)=2n+2$, which is lower than that of the design in fig2b thus, the design of fig 2b is the preferred ripple carry design among those presented so far

FPGA Carry Chain Cell



delay of $2n$ for an n -bit ripple carry chain

50% faster circuit than the original design

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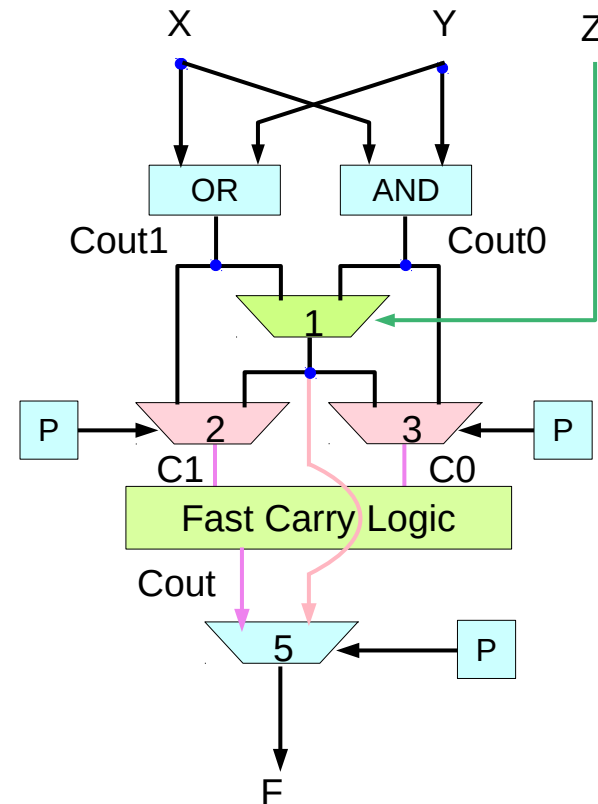
FPGA Carry Chain Cell

various high performance carry chains can be developed based on the carry cell of fig 2c

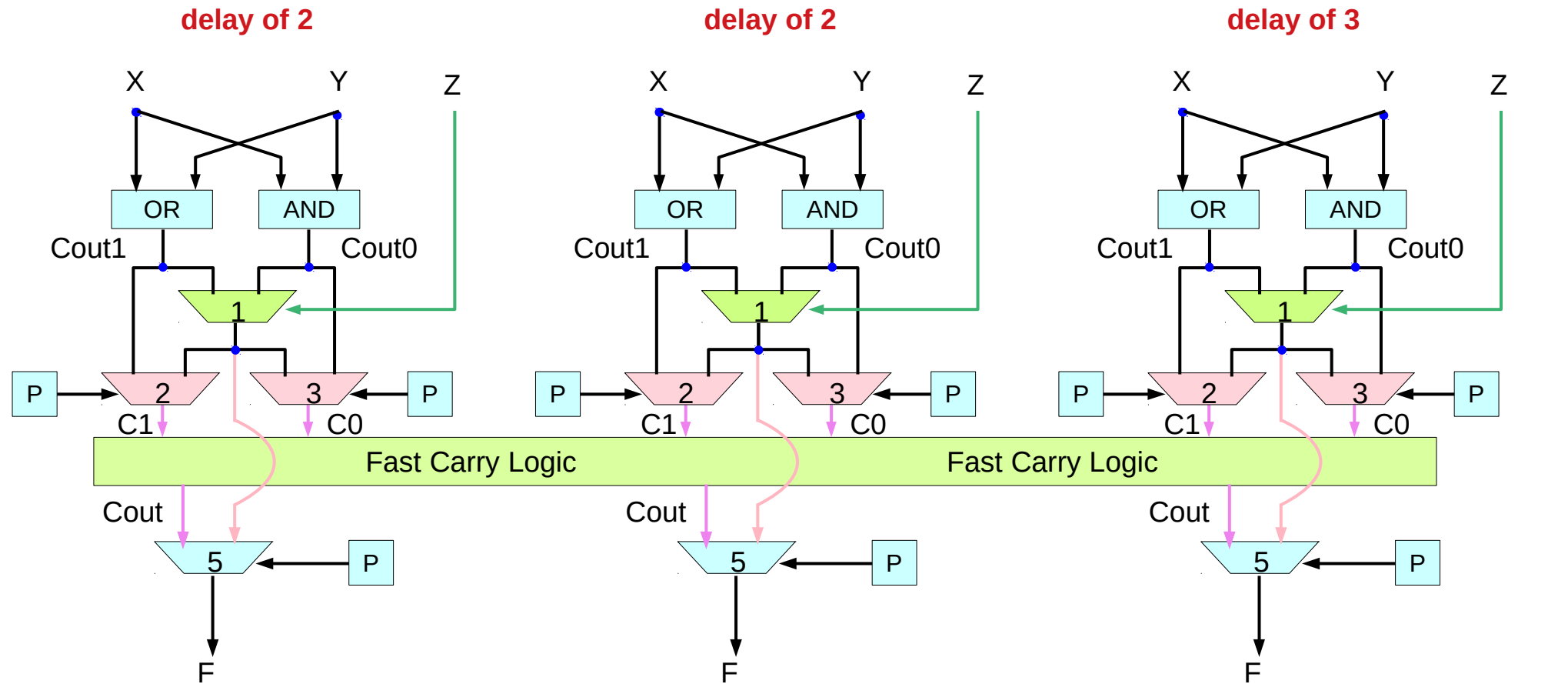
this cell is very similar to that of fig 2b, except that the actual carry chain (mux4) has been replaced by an abstract fast carry logic unit and mux5 has been added

this extra mux5 is present because although some of our faster carry chains will have much faster carry propagation for long carry chains, they incur significant delay for non-carry computations

thus, when the cell is used as a simple normal 3LUT, using inputs X, Y, and Z mux5 allows us to bypass the carry chain by selecting the output of mux1



FPGA Carry Chain Cell



(1 for mux1, 1 for mux2, 1 in mux4)

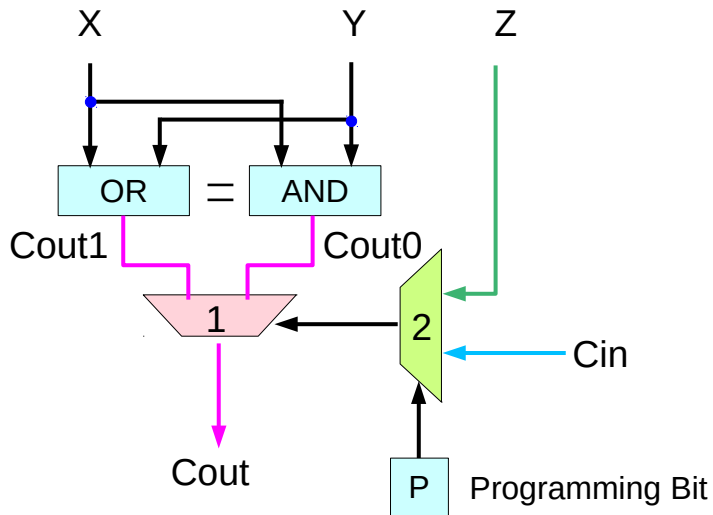
delay of $2n+1$ for an **n-bit** ripple carry chain

1 gate delay slower

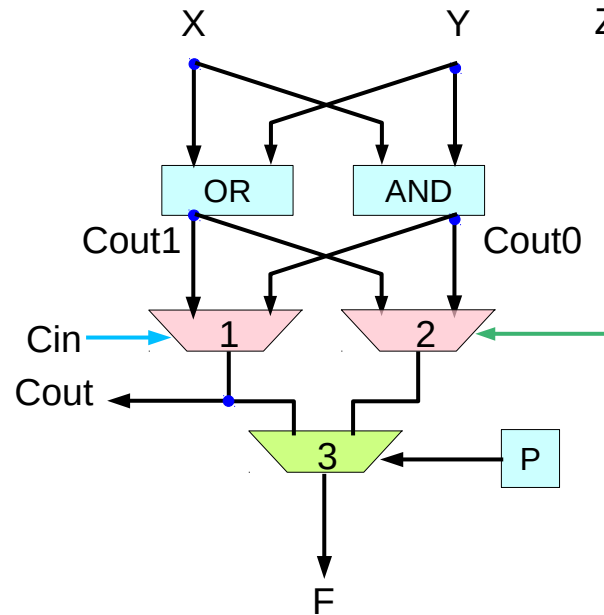
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FPGA Carry Chain Cell

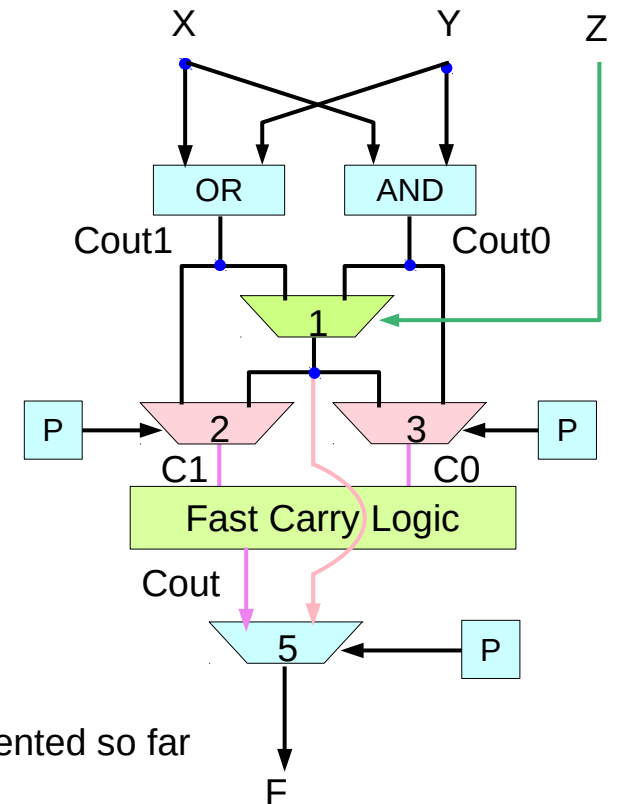
$2n+2$



$2n$



$2n+1$



t
thus, the design of fig 2b is the preferred ripple carry design among those presented so far

FPGA Carry Chain Cell

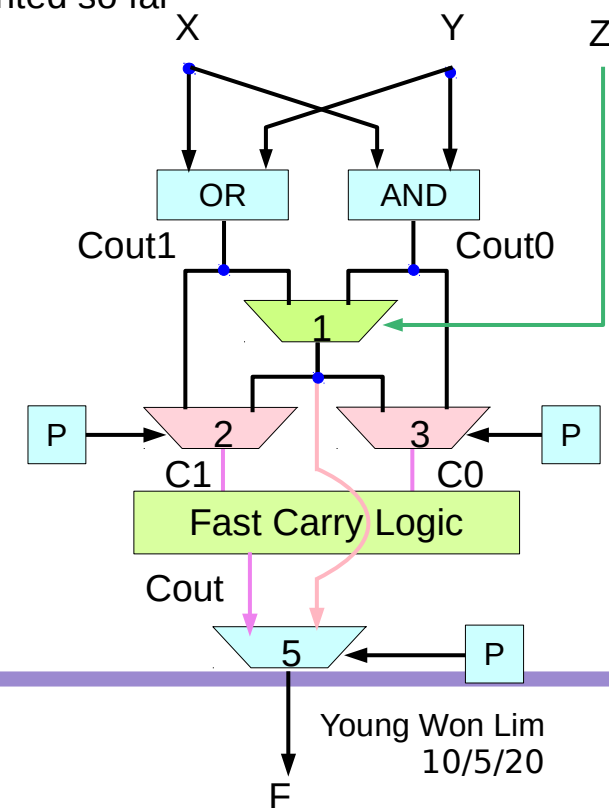
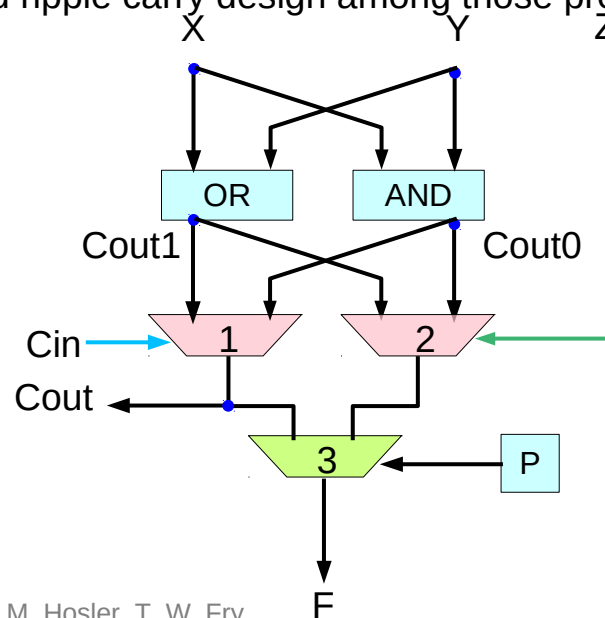
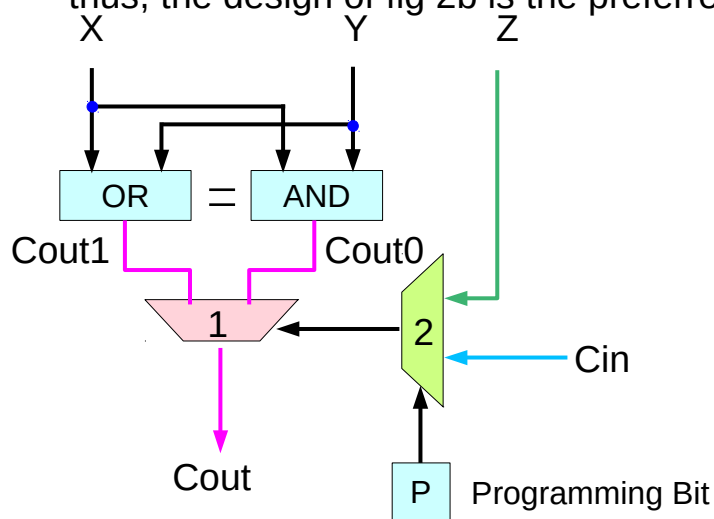
However, carry chains built from this design have a delay of 3 in the first cell (1 in mux1, 1 in mux2, 1 in mux4) and 2 in all other cells in the carry chain, yielding an overall delay of $2n+1$ for an n -bit carry chain.

thus, although this design is 1 gate delay slower than that of fig 2a, it provides the ability to have a carry input to the first cell in a carry chain, something that is important in many computations.

Also, for carry computations that do not need this feature, the first cell in a carry chain built from fig 2b can be configured to bypass mux1, reducing the overall delay to $2n$, which is identical to that of fig 2a.

on the other hand, in order to implement a n -bit carry chain with a carry input, the design of fig 2a requires an additional cell at the beginning of the chain to bring in this input, resulting in a delay of $2(n+1)=2n+2$, which is lower than that of the design in fig 2b

thus, the design of fig 2b is the preferred ripple carry design among those presented so far



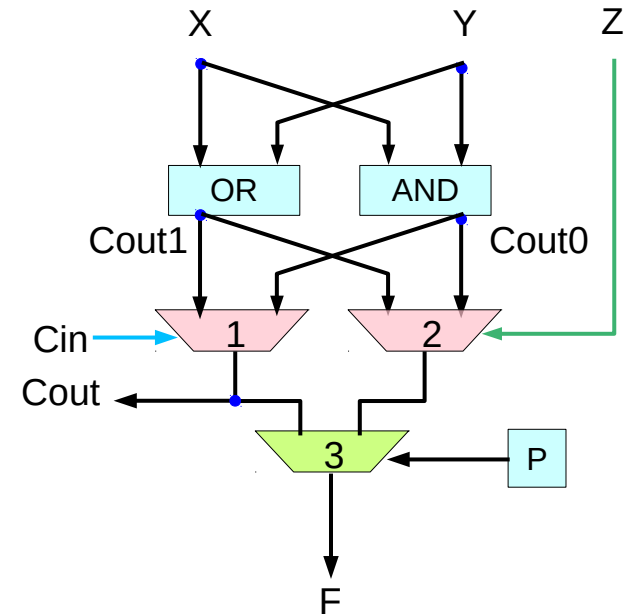
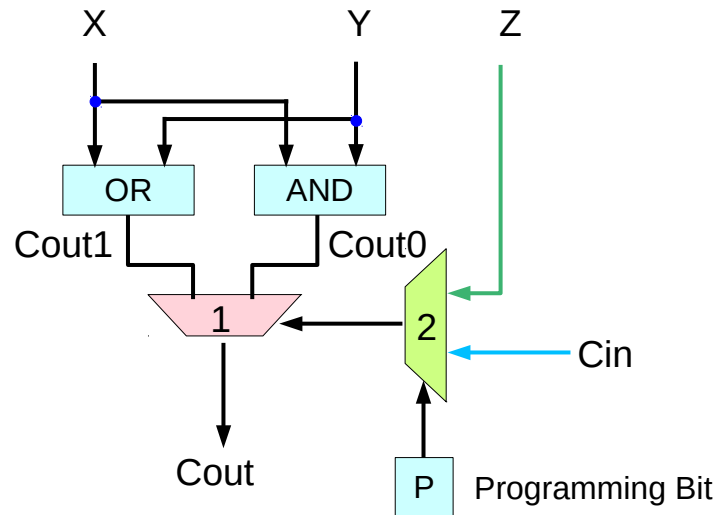
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FPGA Carry Chain Cell

to optimize a ripple carry chain structure for use in FPGAs
while this provides some performance gain over the basis ripple carry scheme
found in many current FPGAs, it is still much slower than what is done in custom logic
There have been tremendous amounts of work done on developing alternative
carry chain scheme that overcome the linear delay growth of ripple carry adders
Although these techniques have not yet been applied to FPGAs,
demonstrate how these advanced adder techniques can be integrated into reconfigurable logic

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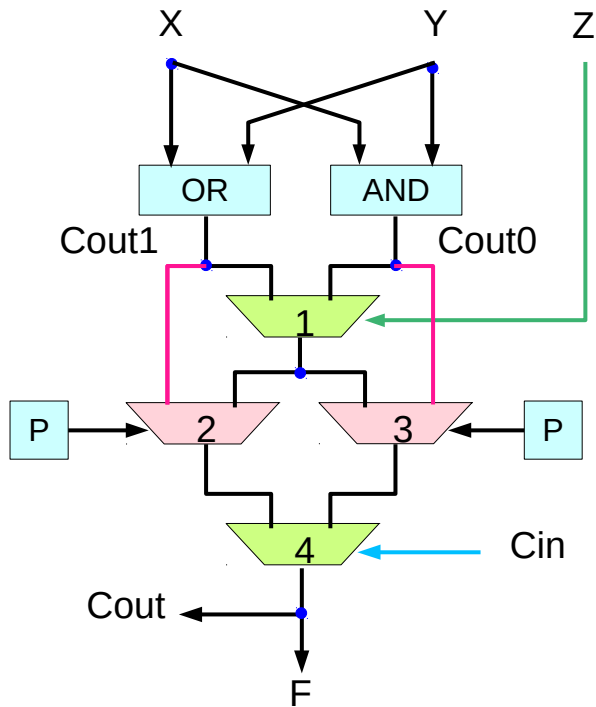
FPGA Carry Chain Cell



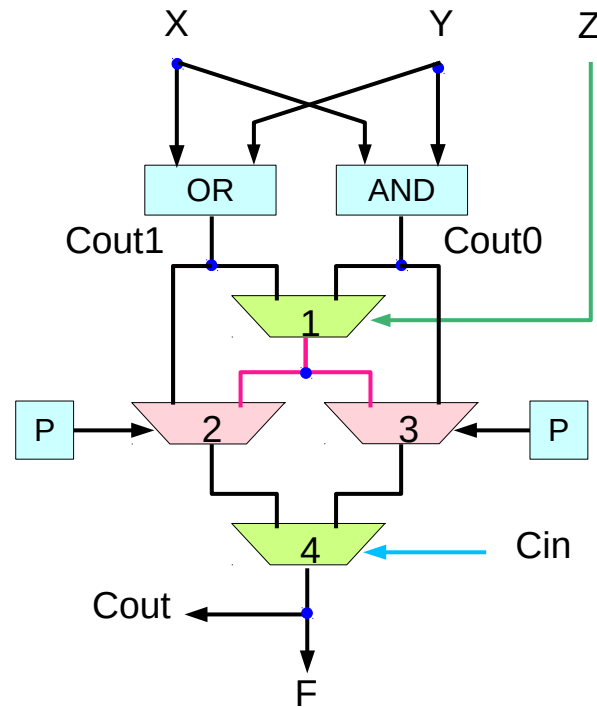
- not logically equivalent
- no longer use the Z input in the first cell since Z is only attached to mux2 and mux 2 does not lead to the carry cells

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FPGA Carry Chain Cell



for cells in the middle of a carry chain
mux2 passes Cout1
mux3 passes Cout0
mux4 receives Cout1 and Cout0
provides a standard ripple carry path.

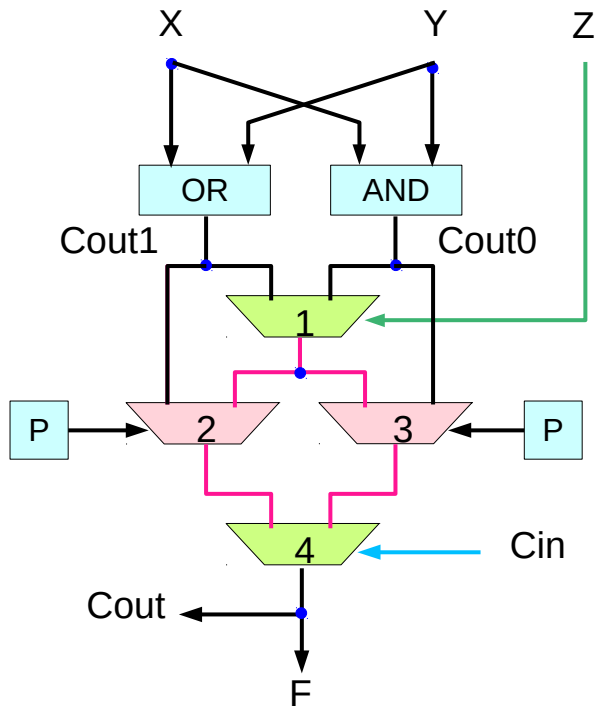


For the first cell in a carry chain
with a carry input (provided by input Z),
mux2 and mux3 both pass the value from mux1

the two main inputs to mux4 are identical
the output of mux4 (Cout) will be the same
as the output of mux1 (ignoring Cin)

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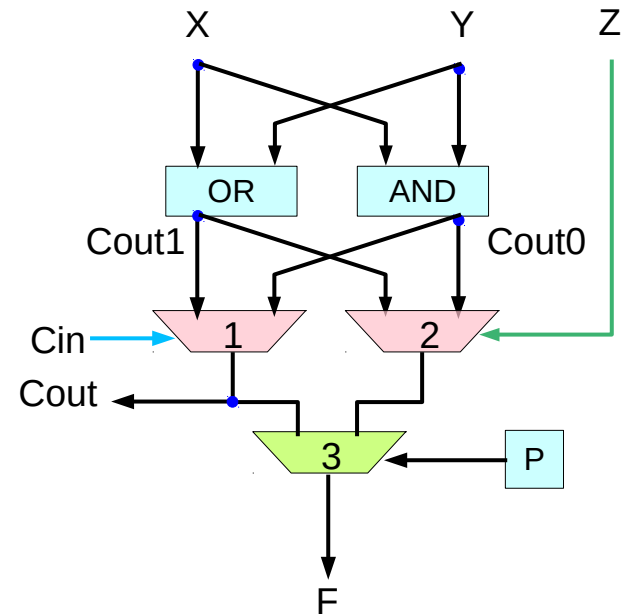
FPGA Carry Chain Cell



mux1's main inputs are driven by two 2-LUTs (OR, AND) controlled by X and Y
mux1 forms a 3-LUT with the other 2-LUTs

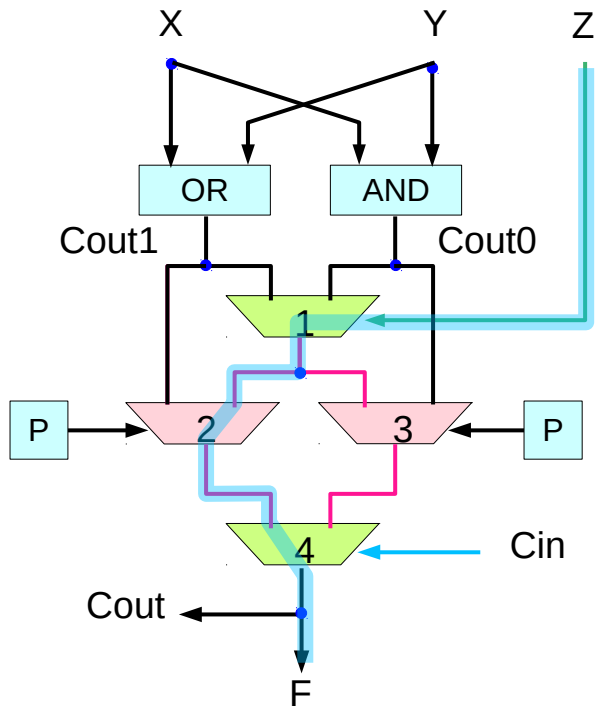
When mux2 and mux3 pass the value from mux1 (Cout1 and Cout2 respectively) the circuit is configured to continue the carry chain

Functionally equivalent



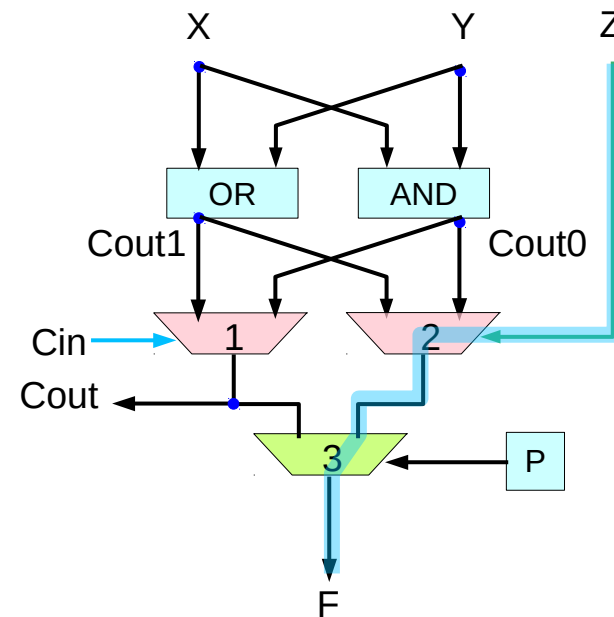
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FPGA Carry Chain Cell



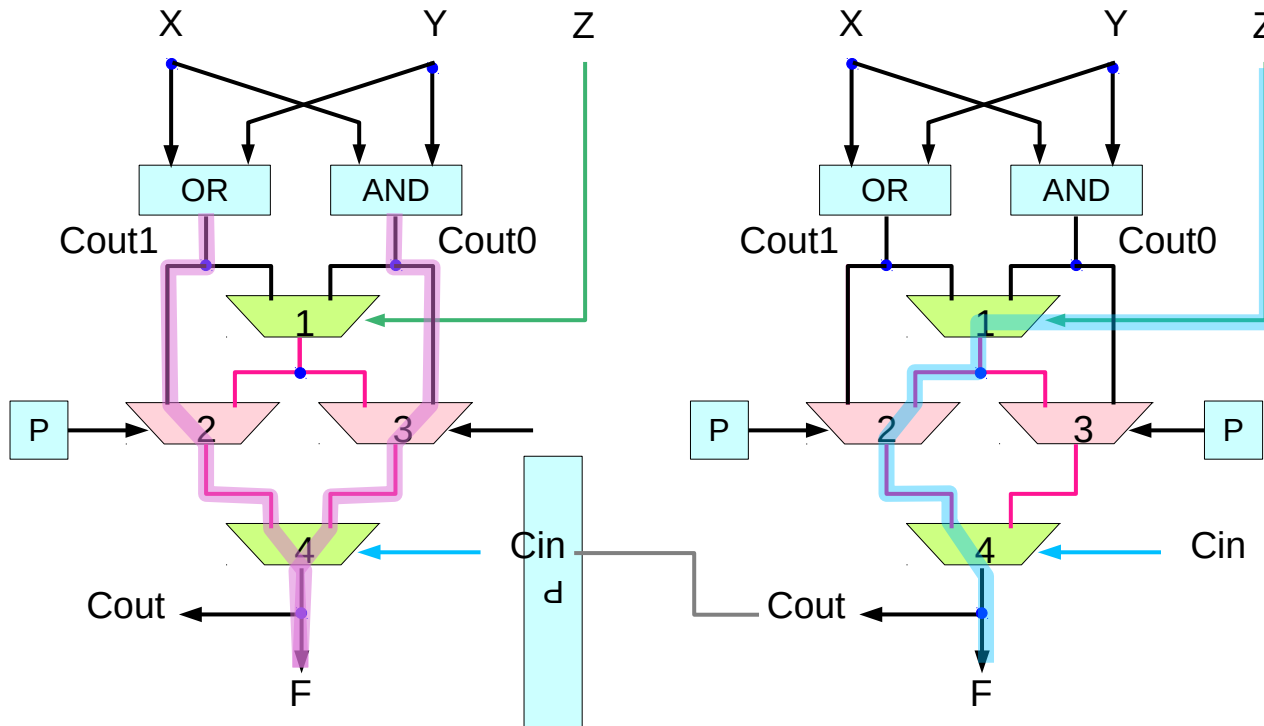
a delay of 3 in the first cell
 (1 in mux1, 1 in mux2, 1 in mux4)
 2 in all other cells in the carry chain
 an total delay of $2n+1$ for an n -bit carry chain

1 gate delay slower than that of fig 2a,
 a carry input to the first cell is enabled



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FPGA Carry Chain Cell



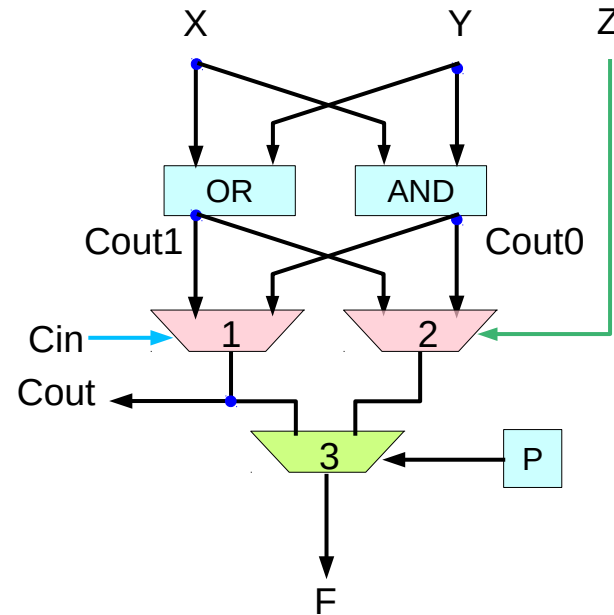
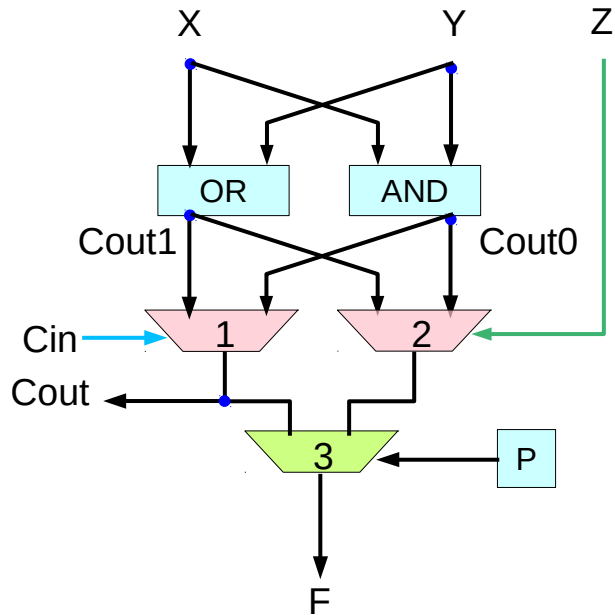
Also, for carry computations that do not need this feature, the first cell in a carry chain built from fig 2b can be configured to bypass mux1, reducing the overall delay to $2n$, which is identical to that of fig2a.

in order to implement a n -bit carry chain with a carry in, the design of fig 2a requires an additional cell at the beginning of the chain to bring in this input, resulting in a delay of $2(n+1)=2n+2$, which is lower than that of the design in fig2b

thus, the design of fig 2b is the preferred ripple carry design among those presented so far

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FPGA Carry Chain Cell

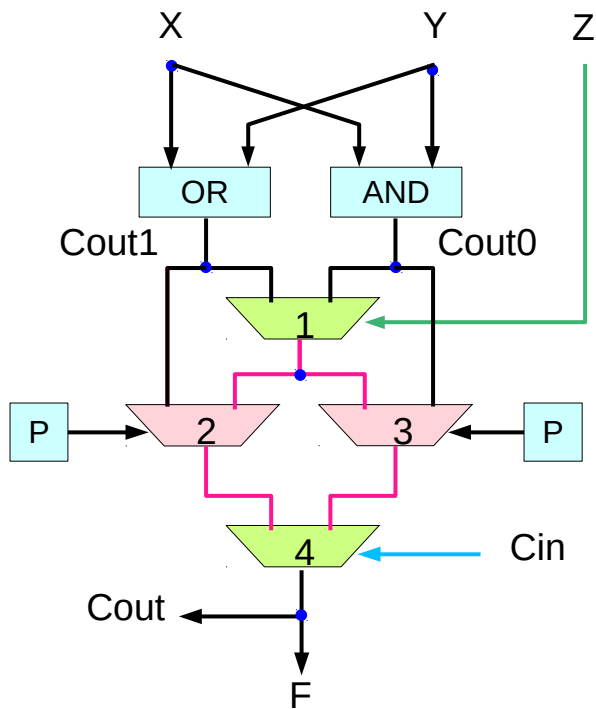


in order to implement a n-bit carry chain with a carry input the design of fig 2a requires an additional cell at the beginning of the chain to bring in this input, resulting in a delay of $2(n+1)=2n+2$, which is lower than that of the design in fig2b

thus, the design of fig 2b is the preferred ripple carry design among those presented so far

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FPGA Carry Chain Cell



a delay of 3 in the first cell
 (1 in mux1, 1 in mux2, 1 in mux4)
 2 in all other cells in the carry chain
 an total delay of $2n+1$ for an n -bit carry chain

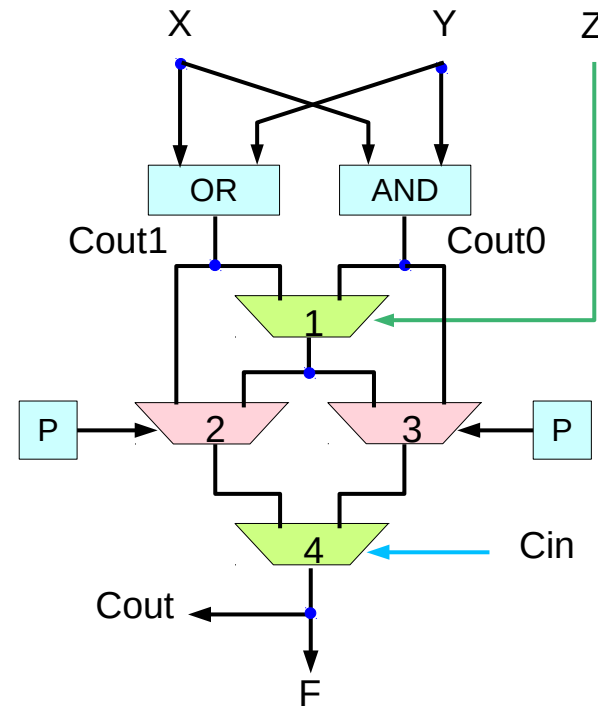
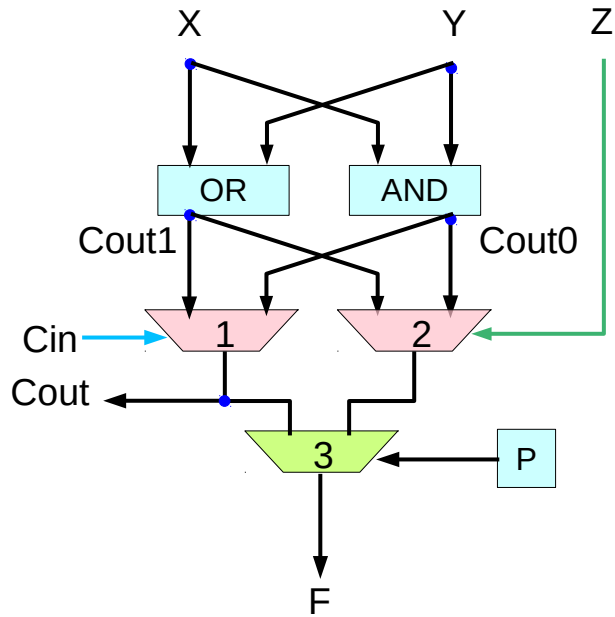
t_1 gate delay slower than that of fig 2a,
 a carry input to the first cell is enabled

Also, for carry computations that do not need this feature,
 the first cell in a carry chain built from fig 2b
 can be configured to bypass mux1,
 reducing the overall delay to $2n$,
 which is identical to that of fig2a.

in order to implement a n -bit carry chain with a carry input,
 the design of fig 2a requires an additional cell
 at the beginning of the chain to bring in this input,
 resulting in a delay of $2(n+1)=2n+2$,
 which is lower than that of the design in fig2b

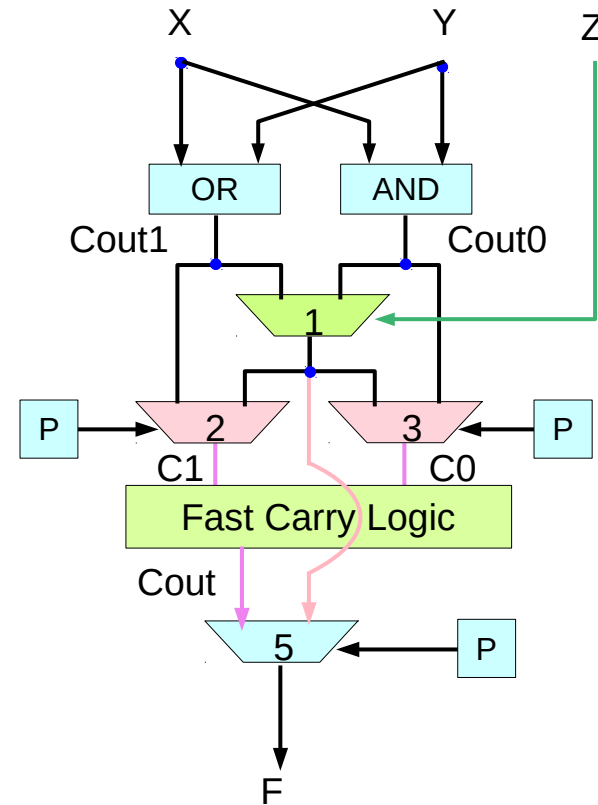
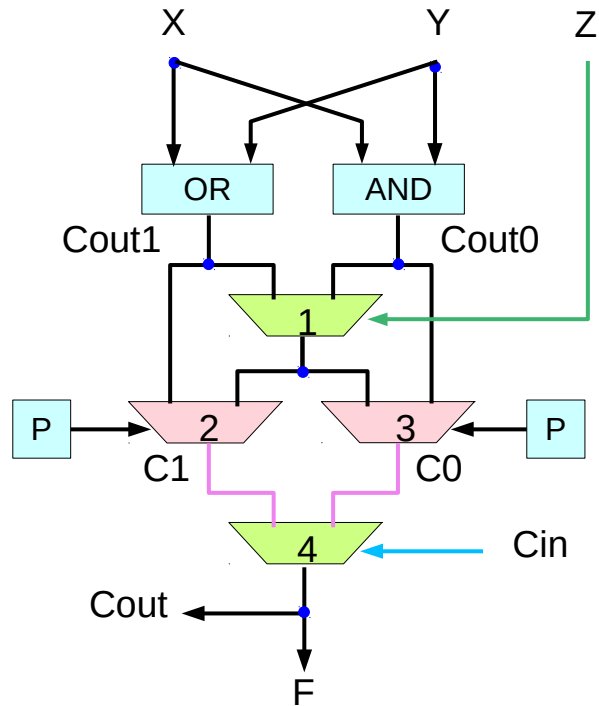
thus, the design of fig 2b is the preferred
 ripple carry design among those presented so far

FPGA Carry Chain Cell



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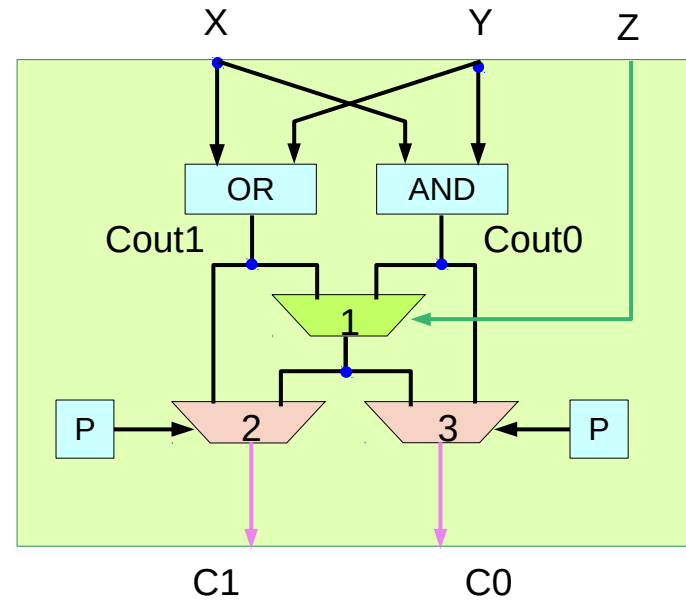
FPGA Carry Chain Cell



$$Cout_i = (Cout_{i-1} \cdot C1_i) + (\overline{Cout_{i-1}} \cdot C0_i)$$

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FPGA Carry Chain Cell



$$Cout_i = (Cout_{i-1} \cdot C1_i) + (\overline{Cout_{i-1}} \cdot C0_i)$$

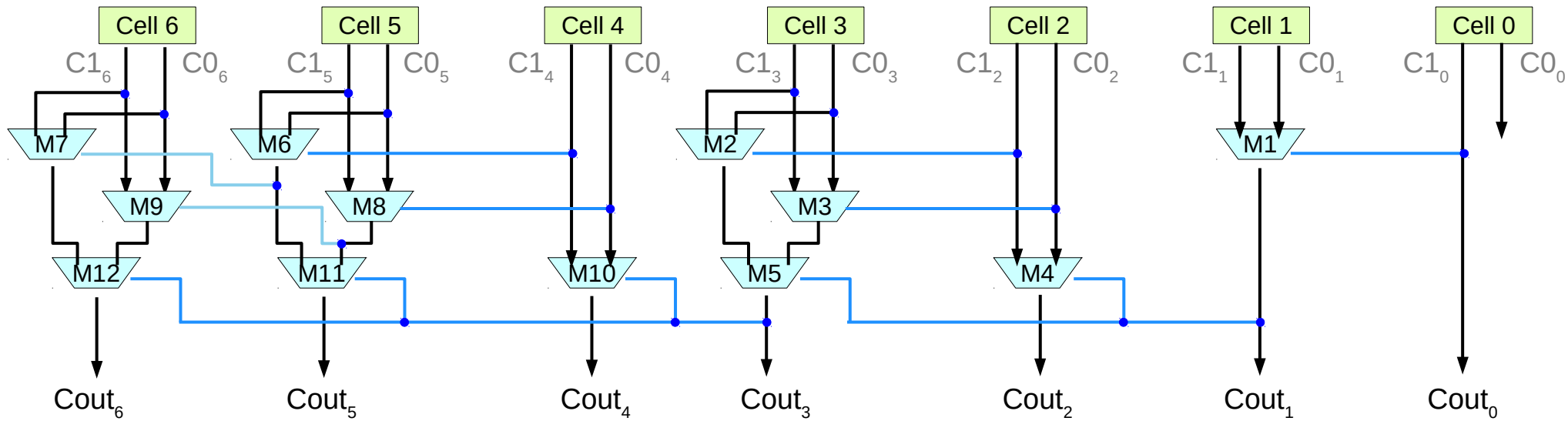
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Fast Carry Logc

Carry Select Adder
Carry Lookahead Adder
 Brent-Kung
Variable Block
Ripple Carry Adder

https://en.wikipedia.org/wiki/Carry-lookahead_adder

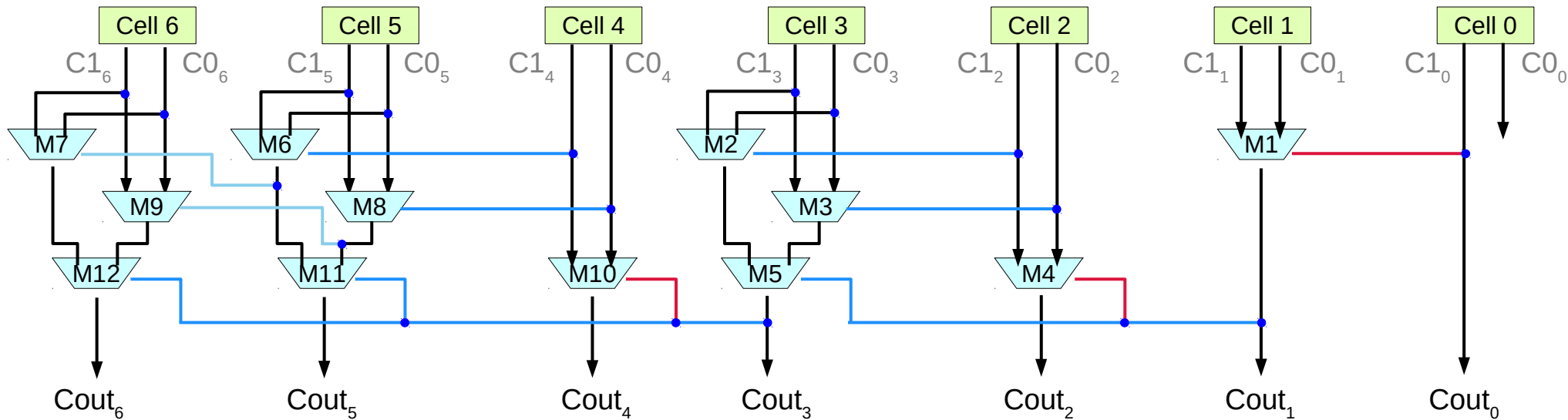
FPGA Carry Chain Cell



$$Cout_i = (Cout_{i-1} \cdot C1_i) + (\overline{Cout_{i-1}} \cdot C0_i)$$

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FPGA Carry Chain Cell



$$Cout_i = (Cout_{i-1} \cdot C1_i) + (\overline{Cout_{i-1}} \cdot C0_i)$$

$$Cout_1 = (Cout_0 \cdot C1_1) + (\overline{Cout_0} \cdot C0_1)$$

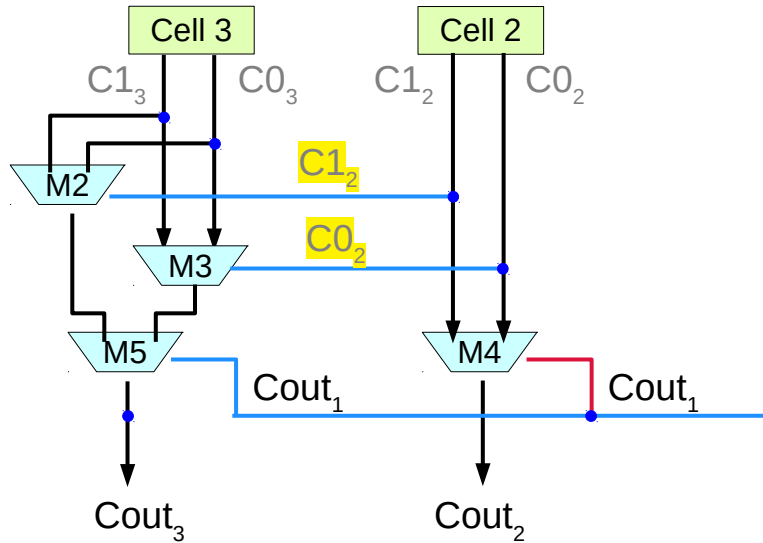
$$Cout_{i+1} = (Cout_i \cdot C1_{i+1}) + (\overline{Cout_i} \cdot C0_{i+1})$$

$$Cout_1 = (C1_0 \cdot C1_1) + (\overline{C1_0} \cdot C0_1)$$

$$Cout_{i+1} = (((Cout_{i-1} \cdot C1_i) + (\overline{Cout_{i-1}} \cdot C0_i)) \cdot C1_{i+1}) + (\overline{((Cout_{i-1} \cdot C1_i) + (\overline{Cout_{i-1}} \cdot C0_i))} \cdot C0_{i+1})$$

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FPGA Carry Chain Cell



$$Cout_i = (Cout_{i-1} \cdot C1_i) + (\overline{Cout_{i-1}} \cdot C0_i)$$

$$Cout_{i+1} = (Cout_i \cdot C1_{i+1}) + (\overline{Cout_i} \cdot C0_{i+1})$$

$$Cout_2 = (Cout_1 \cdot C1_2) + (\overline{Cout_1} \cdot C0_2)$$

$$Cout_3 = (Cout_2 \cdot C1_3) + (\overline{Cout_2} \cdot C0_3)$$

$$= (((Cout_1 \cdot C1_2) + (\overline{Cout_1} \cdot C0_2)) \cdot C1_3)$$

$$+ (((Cout_1 \cdot C1_2) + (\overline{Cout_1} \cdot C0_2)) \cdot C0_3)$$

$$(((Cout_1 \cdot C1_2) + (\overline{Cout_1} \cdot C0_2)) \cdot C1_3)$$

$$= (C1_3 C1_2 Cout_1 + C1_3 C0_2 \overline{Cout_1})$$

$$(C1_3 C1_2 + C0_3 \overline{C1_2}) Cout_1 + (C1_3 C0_2 + C0_3 \overline{C0_2}) \overline{Cout_1}$$

$$(((\overline{Cout_1} \cdot C1_2) + (\overline{Cout_1} \cdot C0_2)) \cdot C0_3)$$

$$= (((\overline{Cout_1} + \overline{C1_2}) \cdot (\overline{Cout_1} + \overline{C0_2})) \cdot C0_3)$$

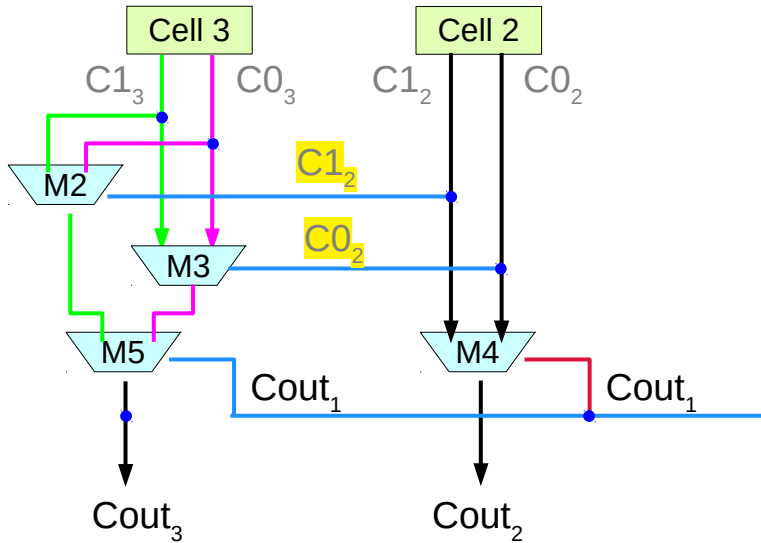
$$= (\overline{Cout_1} \overline{Cout_1} + \overline{C1_2} \overline{Cout_1} + \overline{Cout_1} \overline{C0_2} + \overline{C1_2} \overline{C0_2}) \cdot C0_3$$

$$= (\overline{C1_2} \overline{Cout_1} + \overline{C0_2} \overline{Cout_1}) \cdot C0_3$$

$$= (C0_3 \overline{C1_2} \overline{Cout_1} + C0_3 \overline{C0_2} \overline{Cout_1})$$

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FPGA Carry Chain Cell

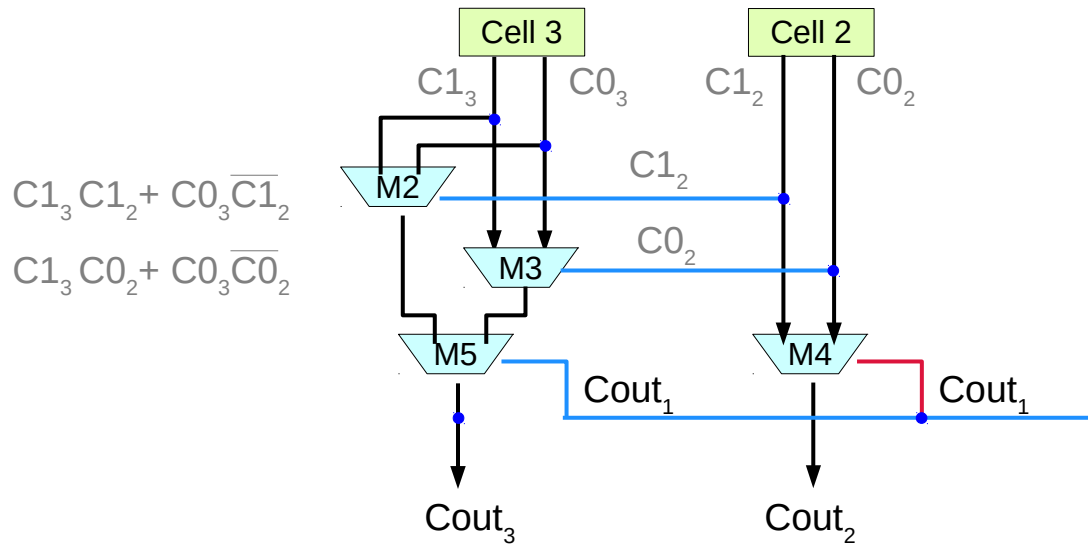


$$\begin{aligned}
 &= (\overline{Cout_1}Cout_1 + \overline{C1_2}Cout_1 + \overline{Cout_1}C0_2 + \overline{C1_2}C0_2) \cdot C0_3 \\
 &= (\overline{C1_2}Cout_1 + \overline{C0_2}Cout_1) \cdot C0_3 \\
 &= (C0_3\overline{C1_2}Cout_1 + C0_3\overline{C0_2}Cout_1)
 \end{aligned}$$

$$(C1_3 C1_2 + C0_3 \overline{C1_2})Cout_1 + (C1_3 C0_2 + C0_3 \overline{C0_2})\overline{Cout_1}$$

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FPGA Carry Chain Cell



$$C1_3 C1_2 + C0_3 \overline{C1_2}$$

$$C1_3 C0_2 + C0_3 \overline{C0_2}$$

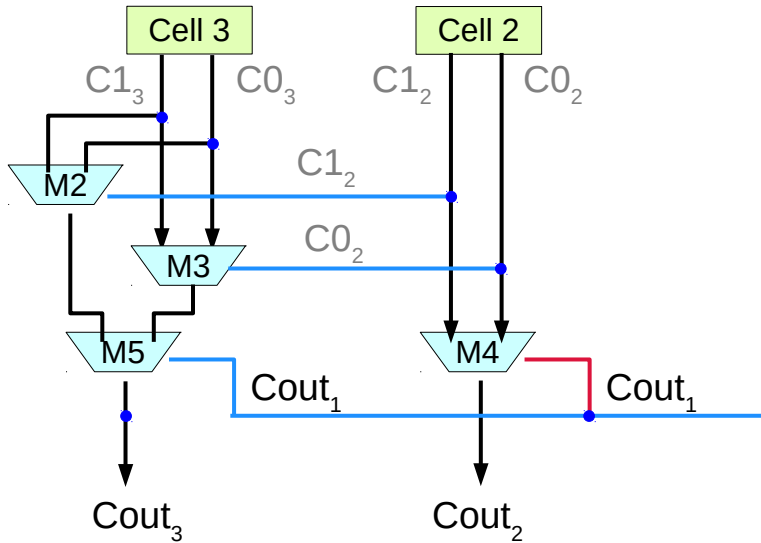
$$(C1_3 C1_2 + C0_3 \overline{C1_2}) Cout_1 + (C1_3 C0_2 + C0_3 \overline{C0_2}) \overline{Cout_1}$$

$$= C1_3 \cdot (C1_2 Cout_1 + C0_2 \overline{Cout_1})$$

$$+ C0_3 \cdot (\overline{C1_2} Cout_1 + \overline{C0_2} \overline{Cout_1})$$

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FPGA Carry Chain Cell



$$Cout_i = (Cout_{i-1} \cdot C1_i) + (\overline{Cout_{i-1}} \cdot C0_i)$$

$$Cout_{i+1} = (Cout_i \cdot C1_{i+1}) + (\overline{Cout_i} \cdot C0_{i+1})$$

$$Cout_{i+1} = [(Cout_{i-1} \cdot C1_i) + (\overline{Cout_{i-1}} \cdot C0_i)] \cdot C1_{i+1} + \overline{[(Cout_{i-1} \cdot C1_i) + (\overline{Cout_{i-1}} \cdot C0_i)]} \cdot C0_{i+1}$$

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References

[1] <http://en.wikipedia.org/>

[2] J-P Deschamps, et. al., “Sunthesis of Arithmetic Circuits”, 2006