

# Lookahead CORDIC Literature

20160723

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Lookahead Technique - CC Kao

Hybrid CORDIC - Wang & Swartzlander (1997)

Low Latency Time CORDIC Algorithms - Timmermann (1992)

Merged CORDIC Algorithms - Wang & Swartzlander (1995)

Merged Scaling Multiplication CORDIC Algorithms - Wang & Swartzlander (1997)

- Takagi (1987)

Redundant and on-line CORDIC - Ercegovic & Lang (1990)

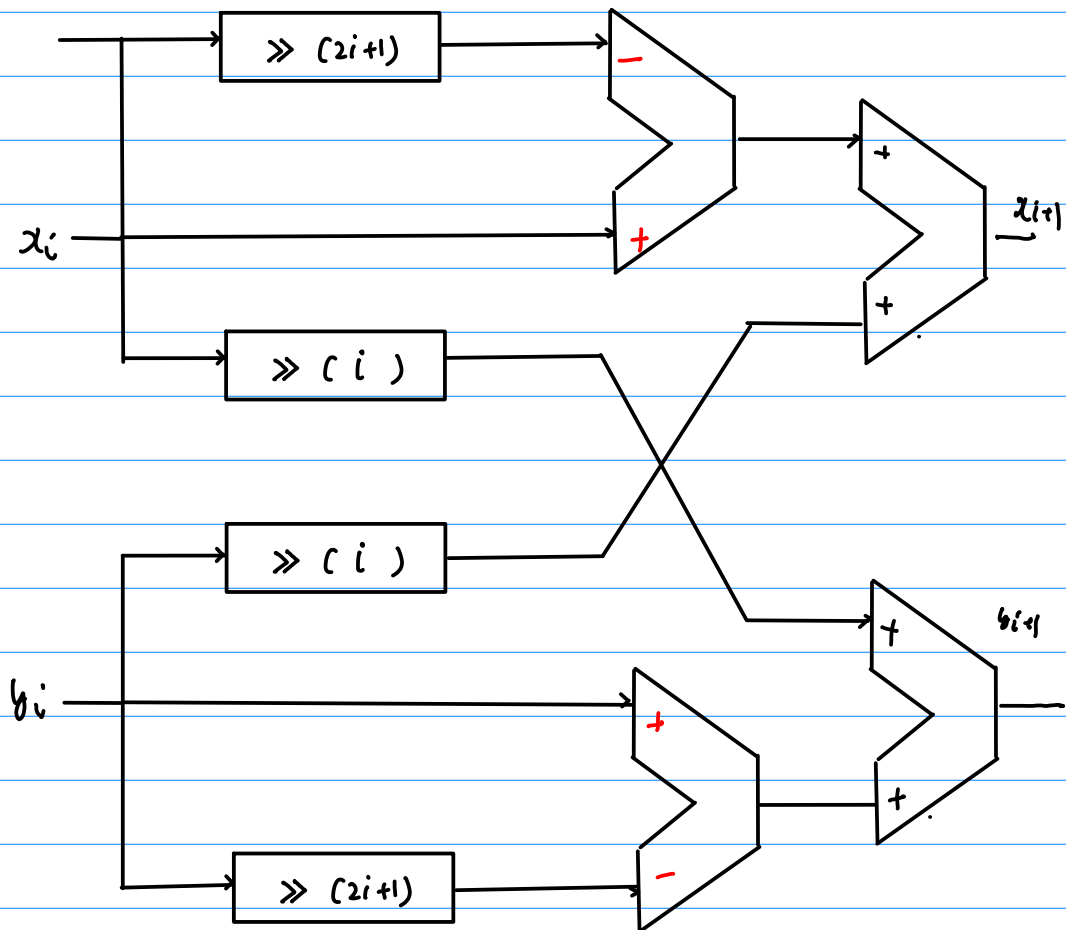
Double Step Branching CORDIC - Phatak (1998)

- Duprat & Muller (1993)

Virtually scaling-free

Maharatna

$$\begin{bmatrix} x' \\ y' \end{bmatrix} = \prod_{i=p}^{b-1} \begin{bmatrix} 1 - 2^{-(2i+1)} & 2^{-i} \\ -2^{-i} & -1 + 2^{-(2i+1)} \end{bmatrix} \begin{bmatrix} x_i \\ y_i \end{bmatrix}$$



$$\begin{bmatrix} x' \\ y' \end{bmatrix} = \begin{bmatrix} 1 - 2^{-2i-1} & \mp(2^{-i} + 2^{-i+1}) \\ \pm(2^{-i} + 2^{-i+1}) & 1 - 2^{-2i-1} \end{bmatrix} \begin{bmatrix} x \\ y \end{bmatrix}$$

$$\begin{bmatrix} x' \\ y' \end{bmatrix} = \prod_{i=p}^{b-1} \begin{bmatrix} 1 - 2^{-(2i+1)} & 2^{-i} \\ -2^{-i} & -1 + 2^{-(2i+1)} \end{bmatrix} \begin{bmatrix} x_i \\ y_i \end{bmatrix}$$

# Lookahead Technique

C.L. Kao

$$x_i = P_{x_i} x_0 - P_{y_i} y_0$$

$$y_i = P_{x_i} y_0 - P_{y_i} x_0$$

$$w_i = w_0 - \sigma_0 \alpha_0 - \dots - \sigma_{i+1} \alpha_{i-1}$$

$$P_{x_1} = 2^{-0} (1)$$

$$P_{y_1} = 2^{-0} (1)$$

$$P_{x_2} = \left(1 - \frac{\sigma_0 \sigma_1}{2}\right)$$

$$P_{y_2} = \left(\sigma_0 + \frac{\sigma_1}{2}\right)$$

$$P_{x_3} = \left(1 - \frac{\sigma_0 \sigma_1}{2} - \frac{\sigma_0 \sigma_2}{4} - \frac{\sigma_1 \sigma_2}{8}\right)$$

$$P_{y_3} = \left(\sigma_0 + \frac{\sigma_1}{2} + \frac{\sigma_2}{4} - \frac{\sigma_0 \sigma_1 \sigma_2}{8}\right)$$

$$P_{x_4} = \left(1 - \frac{\sigma_0 \sigma_1}{2} - \frac{\sigma_0 \sigma_2}{4} - \frac{\sigma_1 \sigma_2}{8} - \frac{\sigma_0 \sigma_3}{8} - \frac{\sigma_1 \sigma_3}{16} - \frac{\sigma_2 \sigma_3}{32} + \frac{\sigma_0 \sigma_1 \sigma_2 \sigma_3}{64}\right)$$

$$P_{y_4} = \left(\sigma_0 + \frac{\sigma_1}{2} + \frac{\sigma_2}{4} - \frac{\sigma_0 \sigma_1 \sigma_2}{8} + \frac{\sigma_3}{8} - \frac{\sigma_0 \sigma_1 \sigma_3}{16} - \frac{\sigma_1 \sigma_2 \sigma_3}{32} - \frac{\sigma_0 \sigma_2 \sigma_3}{64}\right)$$

predict directly the rotation directions  
from the input angle

circuit parallelism

$$\begin{bmatrix} x_1 \\ y_1 \end{bmatrix} = \begin{bmatrix} P_{x_1} & -P_{y_1} \\ P_{y_1} & P_{x_1} \end{bmatrix} \begin{bmatrix} x_0 \\ y_0 \end{bmatrix}$$

$$\begin{bmatrix} x_2 \\ y_2 \end{bmatrix} = \begin{bmatrix} P_{x_2} & -P_{y_2} \\ P_{y_2} & P_{x_2} \end{bmatrix} \begin{bmatrix} x_0 \\ y_0 \end{bmatrix}$$

$$\begin{bmatrix} x_3 \\ y_3 \end{bmatrix} = \begin{bmatrix} P_{x_3} & -P_{y_3} \\ P_{y_3} & P_{x_3} \end{bmatrix} \begin{bmatrix} x_0 \\ y_0 \end{bmatrix}$$

$P_{x_i}$   $P_{y_i}$  precoding matrices of  $x_i$  &  $y_i$   
stored in a LUT

① Wang, Pinn, Swartzlander (1997)

Hybrid CoRnD

the rotation directions

after the first  $m$  iterations

derived from the  $\omega$  remainder — residual angles

at the end of the  $m$ -th iteration

with the conventional scheme

but execution time for  $\omega$  operation

is reduced to  $\frac{1}{3}$

$N$ : significant word size (not including sign)  
 $n$ : the first  $n$  iterations  $n = \left\lceil \frac{N - \log_2 3}{3} \right\rceil$

rotation directions can be computed

$\left\{ \begin{array}{l} \text{in parallel} \\ \text{without error} \end{array} \right.$

Hybrid Radix Sets  $\left\{ \begin{array}{l} \bullet \text{ATR (Arc tangent Radix)} \\ \bullet \text{Radix 2} \end{array} \right.$

an arbitrary angle

- represented by a set of constant angles  $\{\alpha_i\}$

→ acts like radix in a number system

the initial angle

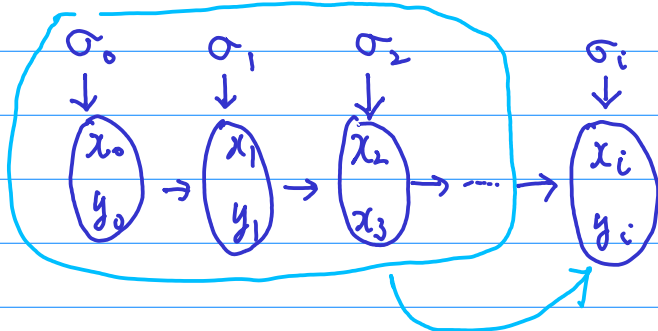
① - represented by a set of arc tangent constants

⇒ ATK (Arc Tangent Radix)

$$\{\alpha_0, \alpha_1, \alpha_2, \dots\} = \{\tan^{-1} 2^0, \tan^{-1} 2^1, \tan^{-1} 2^2, \dots\}$$



Sequential dependence



Sequential dependence

parallelization bottleneck

- ① rotation direction +, -
- ② actual rotation of angles

\* Timmerman 1992

Low Latency Time Correl Algorithms.



⇒ Merged CORDIC {  
- Swartzlander  
- shifter size reduced.

- merging 2 conventional CORDIC iterations  
in the same cycle

⇒ Timmerman 1992  
Low Latency Time CORDIC Algorithms.

- parallel computation of  
rotation directions of a group

⇒ this paper

- partially parallelized

redundant adder 4-to-2

## \* Mixed Hybrid Circular ATR

{ {most significant part}, {least significant part} }

$$= \{ \tan^{-1} 2^0, \tan^{-1} 2^1, \dots, \tan^{-1} 2^{-n+1}, 2^{-n}, \dots, 2^{-N+1} \}$$

parallllism  
exists

## \* Partitioned - Hybrid Circular ATR

{  $\tan^{-1} 2^{-n+1}, 2^{-n}, \dots, 2^{-N+1}$  }

only one iteration

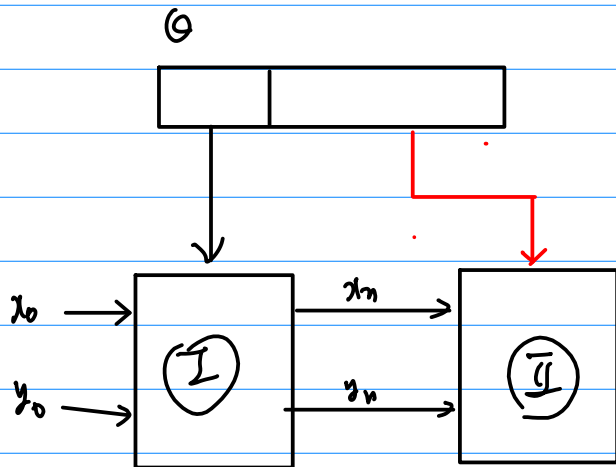
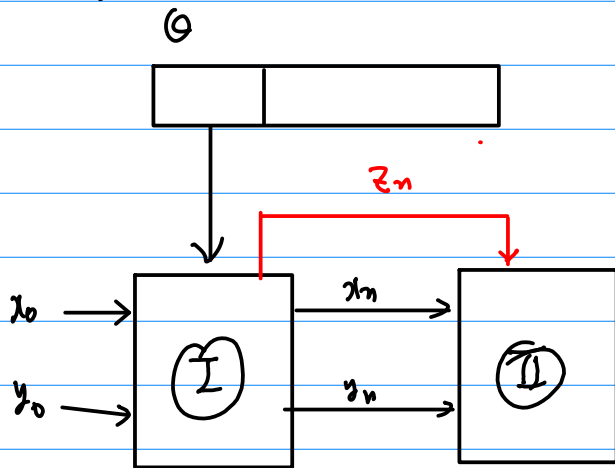
ROM-based LUT.

0 0 ..... 0  $\textcircled{1}$  x x ..... x

any possible angle  $\theta_H$

$$\sigma_{n+1} = \frac{\theta_H}{\tan^{-1} 2^{-n+1}} = \frac{\sum_{i=0}^n \theta_i 2^{-i}}{\tan^{-1} 2^{-n+1}}$$

\* Mixed - Hybrid COPDC



(I)

- ROM

- traditional COPDC

(II)

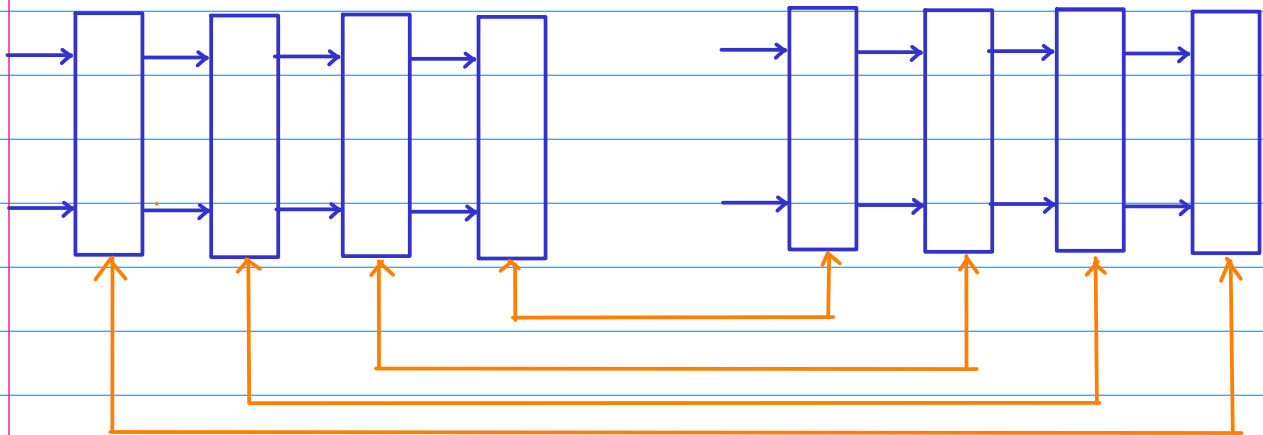
→ Baker's

| $ x_{i+1} $ | $ x_i $ | $x_i$ |
|-------------|---------|-------|
| 0           | 0       | 1     |
| 0           | 1       | 1     |
| 1           | 0       | 1     |
| 1           | 1       | 1     |

\* S. Wang, Swartzlander 1995

### Merged CORDIC Algorithms

↳ approximation error prevents more steps in parallel.



- rearrange the iteration sequence

- merge 2 iterations

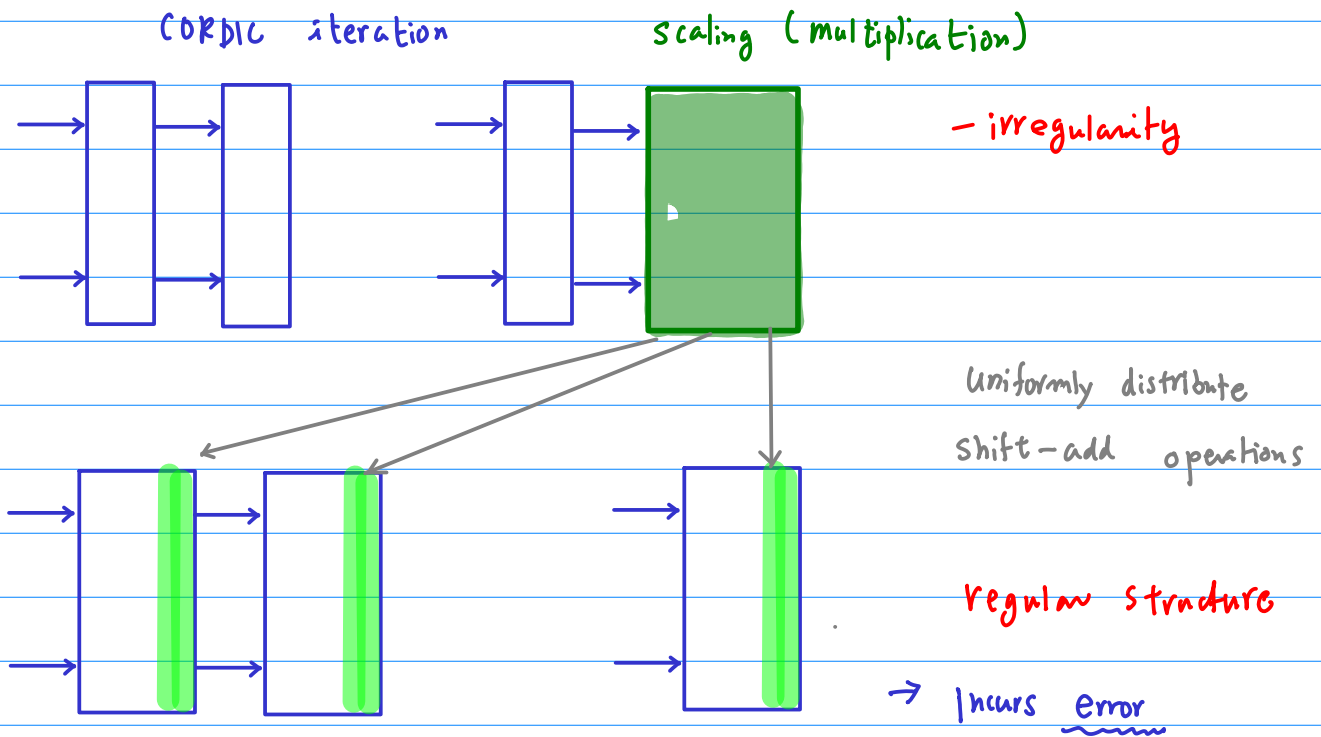
- can reduce the size of barrel shifter by half.

$$\frac{1}{2}$$

\* S. Wang, Swartzlander 1997

Merged Scaling Multiplication CORDIC Algorithms

↳ approximation error prevents more steps in parallel.



the inspection of the  $p$  Most Significant Digits

$\sigma_i = 0$  avoided

the latency time of the inspection  
increases with  $p$

↳ fast carry-dependent adder

$p \leq 4 \sim 5$

allow  $\sigma_i = 0$  valid choice

$$x[j+1] = x[j] + \sigma_j 2^{-j} y[j]$$

$$y[j+1] = y[j] - \sigma_j 2^{-j} x[j]$$

$$w[j] = 2^j y[j]$$

$$2^{-2j} w[j] = 2^{-j} y[j]$$

$$x[j+1] = x[j] + 2^{-2j} w[j]$$

$$y[j+1] =$$





|  $\overline{1}$   $\overline{1}$  |  $\overline{1}$  | |  $\overline{1}$

Lookahead

$$\begin{aligned} &+ \tan^{-1}\left(\frac{1}{2^0}\right) \\ &- \tan^{-1}\left(\frac{1}{2^1}\right) \\ &- \tan^{-1}\left(\frac{1}{2^2}\right) \\ &+ \tan^{-1}\left(\frac{1}{2^3}\right) \\ &- \tan^{-1}\left(\frac{1}{2^4}\right) \\ &+ \tan^{-1}\left(\frac{1}{2^5}\right) \\ &+ \tan^{-1}\left(\frac{1}{2^6}\right) \\ &- \tan^{-1}\left(\frac{1}{2^7}\right) \end{aligned}$$

```
(%i1) a(i) := atan(1/2^i);
(%o1) a(i) := atan(1/2^i)
(%i2) a(0);
(%o2) pi/4
(%i4) %theta : a(0) - a(1) - a(2) + a(3) - a(4) + a(5) + a(6) - a(7);
(%o4) -atan(1/2) - atan(1/4) + atan(1/8) - atan(1/16) + atan(1/32) + atan(1/64) -
atan(1/128) + pi/4
(%i5) float(%theta);
(%o5) 0.17775929681122
```

circular, hyperbolic coordinate system ( $m \neq 0$ )

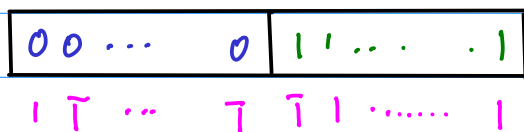
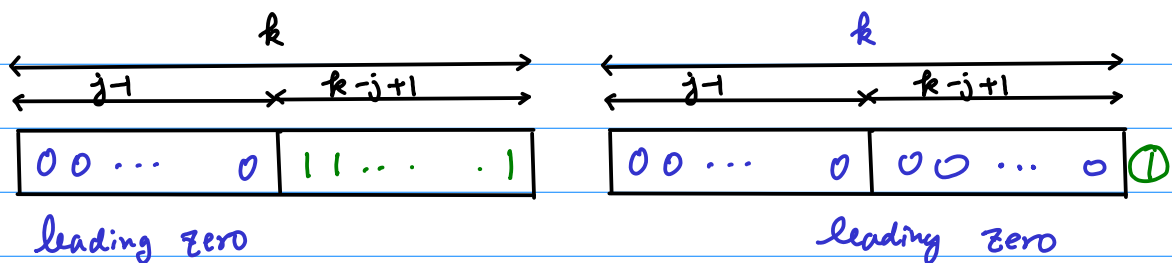
complex conversion  $\leftarrow$  prediction error corrected.

direct conversion not possible

$$\alpha_{m,i} \neq 2^{-i}$$

But  $\alpha_{m,i} \approx 2^{-i}$  for sufficiently large  $i$

can find the upper bound for the prediction error.



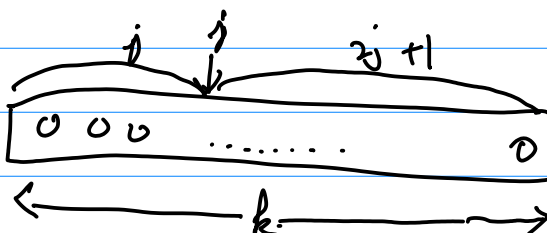
$$\sum_{l=j}^k \left| 2^{-k} - \frac{1}{\sqrt{m}} \tan^{-1}(\sqrt{m} 2^{-s(m,i)}) \right| \leq 2^{-s(m,k)}$$

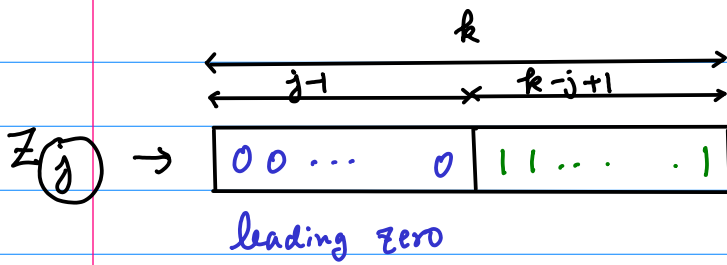
$$s(m, i) = i, \quad m = \pm 1$$

$$j > 0$$

$$k \leq 3j + 1$$

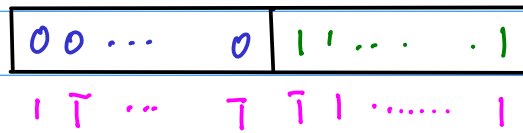
| $j$ | $\max k$ | $k \leq 3j + 1$    |
|-----|----------|--------------------|
| 0   | 1        | $= 3 \cdot 0 + 1$  |
| 1   | 4        | $= 3 \cdot 1 + 1$  |
| 4   | 13       | $= 3 \cdot 4 + 1$  |
| 13  | 40       | $= 3 \cdot 13 + 1$ |
| 40  | 121      | $= 3 \cdot 40 + 1$ |





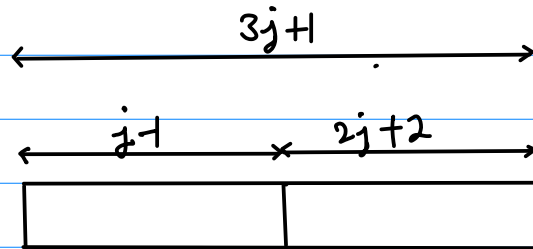
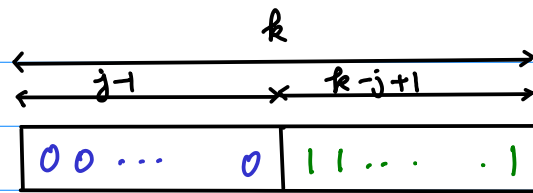
$j$ -th iteration

$Z_j$  assumed to have  
 $(j-1)$  leading zero  
 followed by  $(k-j+1)$  ones



apply bit conversion rule

| $z_{i+1}$ | $z_i$ | $\sigma_i$ |
|-----------|-------|------------|
| 0         | 0     | -1         |
| 0         | 1     | -1         |
| 1         | 0     | 1          |
| 1         | 1     | 1          |



Starting from the  $j$ -th iteration  
 the fully parallel bit conversion rule  
 for determining  $\sigma_i$   
 may be applied  
 for the next  $2j+2$  iterations  
 iteration  $j \sim 3j+1$

followed by a repetition of the last iteration  
 in order to correct any possible errors

\* Timmerman 1992  
Low Latency Time CorePC Algorithms.

redundant addition

booth encoding

$\sigma_i = \pm 1$ , when  $\sigma_i = 0$  stop, freeze iteration  $\rightarrow$  affects scale.

affects  $k_m$ , making it data-dependent.

parallelizing in determining  $\sigma_i$

↑  
prior knowledge of  $\sigma_i$

2 basic rotations in parallel  
in 2 separate modules

2 module perform distinct computation  
only when the algorithm "branches"

2 circular rotations in a single step  
each step involves distinct computation  $\rightarrow$  better.

\* Baker's prediction scheme

p.w. Baker Suggestion for a fast binary sine/cosine generator.

$$(1 + j E_k 2^{-k})$$

$$E_k \in \{-1, +1\}$$

$$\sin X = \text{Im}(U_n)$$

$$\cos X = \text{Re}(U_n)$$

$$U_{k+1} = U_k (1 + j E_k 2^{-k})$$

$$\text{scale } \beta = \prod_{k=0}^{n/2} (1 + 2^{-k})^{-1/2}$$

$$0 \leq k \leq n \quad n: \text{the bit length of } X$$

When  $E_k$ 's can be predicted

$$\text{since } \tan^{-1}(2^{-k}) = 2^{-k} - \frac{2^{-3k}}{3} + \dots$$

```
(%i21) taylor(atan(x), x, 0, 16);
```

```
(%o21)/T/ x -  $\frac{x^3}{3} + \frac{x^5}{5} - \frac{x^7}{7} + \frac{x^9}{9} - \frac{x^{11}}{11} + \frac{x^{13}}{13} - \frac{x^{15}}{15} + \dots$ 
```

→ given  $X_r$  with  $(l-1)$  leading zeros



$$X_k = \pm 0.00 \dots 0 x_{k+1} x_{k+2} \dots x_{2k} \dots x_{3k} x_{3k+1} \dots$$

by setting  $x_i = E_i \quad 1 \leq i \leq 3k$

$$X_0 = X, \quad U_0 = \beta + j0 \quad \beta = a.b0 \dots$$

$$X_{k+1} = X_k - E_k T_k \quad X_{k+1} \rightarrow 0 \quad 0 \leq k \leq n$$

$$T_k = \tan^{-1}(2^{-k})$$

guaranteed  $X_{3k+1} \rightarrow (3k-1)$  leading zeros

the  $3k$ -th bit  $X_{3k+1} \rightarrow +1 (-1)$

When the 3rd order term of  $\tan^{-1}(2^{-k})$

produces **carry (borrow)**

out of the  $(3k+1)$ st bit position

into the  $(3k)$ -th bit position

| $ x_{i-1} $ | $ x_i $ | $\dot{x}_i$ |
|-------------|---------|-------------|
| 0           | 0       | 1           |
| 0           | 1       | 1           |
| 1           | 0       | 1           |
| 1           | 1       | 1           |

② Phatak, D.S. (1998)

Double Step Branching CORDIC

IEEE Trans. on Computers, 47, 589-602

2 rotations are executed in a single step

more complicated  $\omega$ -data path

several most significant digits are examined

Duprat & Muller 1993

CORDIC, New Results Fast VLSI Implementation.

Comments on Duprat and Muller's  
Branching CORDIC

③ Kwak, Choi, Swartzlander 2000

High SpeedCORDIC based on the overlapped Architecture

Journal of VLSI Signal Processing 25, 167-178

1st rotation directions are predicted

based on the approximation of

the binary angle input

# \* Application Specific Processor

edited by Swartzlander

→ [books.google.com](https://books.google.com)

ch2. Modelling the power consumption

ch3. Fault tolerant Arithmetic

ch4. Low Power Digital Multiplier

ch5. Unified View of COPIC processor design

ch6. Multi-dimensional Systolic Arrays → Hypercube Lim.

↳ DCT, DFT.





