Spectrum Representation (2B)

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ω_{s} and ω_{0}



Fourier Transform Types

Continuous Time Fourier Series

$$C_{k} = \frac{1}{T} \int_{0}^{T} x(t) e^{-jk\omega_{0}t} dt \quad \Longleftrightarrow x(t) = \sum_{k=-\infty}^{+\infty} C_{k} e^{+jk\omega_{0}t}$$

Discrete Time Fourier Series

$$\gamma[\mathbf{k}] = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j\mathbf{k}\hat{\omega}_0 n} \quad \longleftrightarrow \ x[n] = \sum_{k=0}^{N-1} \gamma[k] e^{+jk\hat{\omega}_0 n}$$

Continuous Time Fourier <u>Transform</u>

$$X(j\omega) = \int_{-\infty}^{+\infty} x(t) \ e^{-j\omega t} dt \qquad \Longleftrightarrow \ x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) \ e^{+j\omega t} d\omega$$

Discrete Time Fourier Transform

$$X(j\hat{\boldsymbol{\omega}}) = \sum_{n=-\infty}^{+\infty} x[n] e^{-j\hat{\boldsymbol{\omega}}n} \qquad \longleftrightarrow \quad x[n] = \frac{1}{2\pi} \int_{-\pi}^{+\pi} X(j\hat{\boldsymbol{\omega}}) e^{+j\hat{\boldsymbol{\omega}}n} d\hat{\boldsymbol{\omega}}$$

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Computation at $k\omega_0$



Computations using DFT



CTFTAperiodic x(t)
$$X(j\omega) = \int_{-\infty}^{+\infty} x(t) e^{-j\omega t} dt$$
 $X(jk\omega_0) \approx T_s DFT\{x(nT_s)\}$ $\omega \leftarrow k\omega_0$ $(\omega) = k\hat{\omega}_0 f_s = k\left(\frac{2\pi}{NT_s}\right) rad/sec$

DTFS Periodic x[n] $\gamma[k] = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-jk\hat{\omega}_0 n}$ $\gamma[k] = \frac{1}{N} DFT\{x[n]\}$ $k\hat{\omega}_0$ $(a) k\omega_0 = k\hat{\omega}_0 f_s = k \left(\frac{2\pi}{NT_s}\right) rad/sec$

DTFT Aperiodic x[n]

$$X(j\hat{\omega}) = \sum_{n=-\infty}^{+\infty} x[n] e^{-j\hat{\omega}n}$$

$$X(jk\hat{\omega}_{0}) \approx DFT\{x[n]\}$$

$$\hat{\omega} \leftarrow k\hat{\omega}_{0}$$

$$(a) k\omega_{0} = k\hat{\omega}_{0}f_{s} = k\left(\frac{2\pi}{NT_{s}}\right) rad/sec$$

Computations using DFT



FFT Amplitude and Power Spectrum

Two-Sided Amplitude Spectrum

$$A_{k} = \frac{1}{N} |X[k]| \qquad (V/Hz^{-1/2})$$
$$= \frac{1}{N} \sqrt{\Re^{2} \{X[k]\} + \Im^{2} \{X[k]\}}$$

$$k = 0, 1, 2, \cdots, N/2, N/2 + 1, \cdots, N - 1$$

One-Sided Amplitude Spectrum

$$\bar{A}_0 = \frac{1}{N} |X[0]| \qquad k = 0$$
$$\bar{A}_k = \frac{2}{N} |X[k]| \qquad k = 1, 2, \dots, N/2$$

Frequency Bin

$$f = k \frac{f_s}{N}$$

Two-Sided Power Spectrum

$$P_{k} = \frac{1}{N^{2}} |X[k]|^{2} \qquad (V^{2}/Hz^{-1})$$
$$= \frac{1}{N^{2}} (\Re^{2} \{X[k]\} + \Im^{2} \{X[k]\})$$

$$k = 0, 1, 2, \cdots, N/2, N/2 + 1, \cdots, N - 1$$

One-Sided Power Spectrum

$$\bar{P}_{0} = \frac{1}{N^{2}} |X[0]|^{2} \quad k = 0$$
$$\bar{P}_{k} = \frac{2}{N^{2}} |X[k]|^{2} \quad k = 1, 2, \dots, N/2$$

Frequency Bin

$$f = k \frac{f_s}{N}$$

Spectrum Representation (2B)

FFT Amplitude and Phase Spectrum

Two-Sided Amplitude Spectrum

$$A_{k} = \frac{1}{N} |X[k]|$$
$$= \frac{1}{N} \sqrt{\Re^{2} \{X[k]\} + \Im^{2} \{X[k]\}}$$

$$k = 0, 1, 2, \cdots, N/2, N/2 + 1, \cdots, N - 1$$

Two-Sided Phase Spectrum

$$\phi_k = \tan^{-1} \left(\frac{\Im \{X[k]\}}{\Re \{X[k]\}} \right)$$

$$k = 0, 1, 2, \dots, N/2, N/2 + 1, \dots, N - 1$$

Frequency Bin

$$f = k \frac{f_s}{N}$$

Spectrum Representation (2B)

CTFS and Power Spectrum

Two-Sided Power Spectrum

$$\frac{1}{N^2} |X[k]|^2 = |C_k|^2 = \frac{1}{4} (a_k^2 + b_k^2) = \frac{1}{4} |g_k|^2$$

Single-Sided Power Spectrum

$$\frac{2}{N^2} |X[k]|^2 = 2 |C_k|^2 = \frac{1}{2} |g_k|^2 = |g_{k,rms}|^2$$

$$C_{k} = \frac{1}{2}g_{k} e^{ij\phi_{k}} \quad (k > 0)$$

$$C_{k} = \frac{1}{2}g_{-k} e^{-j\phi_{k}} \quad (k < 0)$$

$$x(t) = g_{0} + \sum_{k=1}^{\infty} g_{k} \cos(k\omega_{0}t + \phi_{k})$$

 g_k each sinusoid's amplitude $g_{k,rms}$ each sinusoid's amplitude rms value

Average Power and Total Energy



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Parseval's Theorem for DFT



Periodogram as a frequency domain samples



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Parseval's Theorem



Approximate CTFS Parseval's Theorem



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DTFS Parseval's Theorem



Approximate CTFT Parseval's Theorem



Approximate **DTFT** Parseval's Theorem



Average Power and Total Energy

Periodic Signals	Aperiodic Signals
Average Power	Total Energy



Parseval's Theorem

Periodic Signals	Aperiodic Signals
Average Power	Total Energy

Continuous Time	
Discrete Time	

CTFS Average Power	CTFT Total Energy
$\frac{1}{T}\int_{T} x(t) ^{2}dt = \sum_{k=-\infty}^{+\infty} C_{k} ^{2}$	$\int_{T} x(t) ^{2} dt = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) ^{2} d\omega$
DTFS Average Power	DTFT Total Energy
$\frac{1}{N}\sum_{n=0}^{N-1} x[n] ^2 = \sum_{k=0}^{N-1} y[k] ^2$	$T_{s} \sum_{n=0}^{N-1} x[n] ^{2} = \frac{T_{s}}{2\pi} \int_{2\pi} X(j\hat{\omega}) ^{2} d\hat{\omega}$

Average Power and Total Energy

Periodic Signals	Aperiodic Signals
Average Power	Total Energy

Average Power CTFS CTFT **Total Energy Continuous** Time $\cdot T$ $\frac{1}{T}\int_{T}|x(t)|^{2}dt$ $\int |x(t)|^2 dt$ DTFS DTFT **Average Power Total Energy Discrete** Time **Total Energy Average Power**

DFT Approximation

Periodic Signals	Aperiodic Signals
Average Power	Total Energy

CTFSAverage PowerCTFTTotal Energy $\frac{1}{N^2} \sum_{k=0}^{N-1} |X[k]|^2 = \frac{1}{N} \sum_{n=0}^{N-1} |x[n]|^2$ $\frac{T_s}{N} \sum_{k=0}^{N-1} |X[k]|^2 = T_s \sum_{n=0}^{N-1} |x[n]|^2$ DTFSAverage PowerDTFTTotal Energy $\frac{1}{N^2} \sum_{k=0}^{N-1} |X[k]|^2 = \frac{1}{N} \sum_{n=0}^{N-1} |x[n]|^2$ $\frac{T_s}{N} \sum_{k=0}^{N-1} |X[k]|^2 = T_s \sum_{n=0}^{N-1} |x[n]|^2$

Continuous

Time

Discrete

Time



Fourier Series Coefficients

	Periodic Signals	Aperiodic Signals
Frequency Spacing	$\omega_0 = \frac{2\pi}{NT_s}$	$\omega_0 = \frac{2\pi}{NT_s} \left(\hat{\omega}_0 = \frac{2\pi}{N}\right)$
Two Sided F.S. Coefficient	$\frac{1}{N}X[k] = C_k$	$\frac{T_0}{N} X[k] = T_s X[k] = X(jk\omega_0)$
Frequency Bin	$k\omega_0 = k \left(\frac{2\pi}{NT_s}\right)$	$k\omega_0 = k\hat{\omega}_0 f_s = k \left(\frac{2\pi}{NT_s}\right)$
	\mathbf{C}_{i} will be spread	over f_{μ} $(C_{\mu} / f_{\mu} = C_{\mu} T_{\mu})$



One-sided Fourier Series Coefficients

	Periodic Signals	Aperiodic Signals
Frequency Spacing	$\omega_0 = \frac{2\pi}{NT_s} = \frac{2\pi}{T_0}$	$\omega_0 = \frac{2\pi}{NT_s} \left(\hat{\omega}_0 = \frac{2\pi}{N}\right)$
Two Sided F.S. Coefficient	$\frac{1}{N}X[k] = C_k$	$\frac{T_0}{N} X[k] = X(jk\omega_0)$
One Sided F.S. Coefficient	$\frac{1}{N}X[k] k=0, \ \frac{N}{2}$	$\frac{T_0}{N} X[k] \qquad k=0, \ \frac{N}{2}$
	$\frac{2}{N}X[k] k=1,\cdots,\frac{N}{2}-1$	$\frac{2I_0}{N}X[k] \qquad k=1,\cdots,\frac{N}{2}-1$
Frequency Bin	$k\omega_0 = k\left(\frac{2\pi}{NT_s}\right)$	$k\omega_0 = k\hat{\omega}_0 f_s = k \left(\frac{2\pi}{NT_s}\right)$
	Average Power	Total Energy

Spectral Density Functions



Using Periodograms

	Periodic Signals	Aperiodic Signals
Two Sided F.S. Coefficient	$\frac{1}{N}X[k] = C_k$	$\frac{T_0}{N} X[k] = X(jk\omega_0)$
Parseval's Theorem	$\sum_{k=0}^{N-1} C_k ^2 = \frac{1}{T} \int_T x(t) ^2 dt$	$\frac{1}{2\pi}\int_{-\infty}^{+\infty} X(j\omega) ^2d\omega=\int_{-\infty}^{+\infty} x(t) ^2dt$
		$\cdot \omega_0 \left(= \frac{2 \pi}{T_0} \right)$
Approximation By DFT's	$\sum_{k=0}^{N-1} \left \frac{X[k]}{N} \right ^2 = \frac{T_s}{T} \sum_{n=0}^{N-1} x[n] ^2$	$\frac{1}{NT_s} \sum_{k=0}^{N-1} T_s X[k] ^2 = T_s \sum_{n=0}^{N-1} x[n] ^2$
	$\frac{1}{N}\sum_{k=0}^{N-1} \left\{ \frac{1}{N} X[k] ^2 \right\} = \frac{1}{N}\sum_{n=0}^{N-1} x[n] ^2$	$T_{s}\sum_{k=0}^{N-1} \left\{ \frac{1}{N} X[k] ^{2} \right\} = T_{s}\sum_{n=0}^{N-1} x[n] ^{2}$
	Averaging Operation	Integrating Operation
Using Periodograms	$\frac{1}{N} \sum_{k=0}^{N-1} P_{xx}[k] = \frac{1}{N} \sum_{n=0}^{N-1} x[n] ^2$	$T_{s} \sum_{k=0}^{N-1} P_{xx}[k] = T_{s} \sum_{n=0}^{N-1} x[n] ^{2}$
	Average Power	Total Energy
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Using Periodograms

	Periodic Signals	Aperiodic Signals
Two Sided F.S. Coefficient	$\frac{1}{N}X[k] = C_k$	$\frac{T_0}{N}X[k] = X(jk\omega_0)$
Parseval's Theorem	$\sum_{k=0}^{N-1} C_k ^2 = \frac{1}{T} \int_T x(t) ^2 dt$	$\frac{1}{2\pi}\int_{-\infty}^{+\infty} X(j\omega) ^2d\omega=\int_{-\infty}^{+\infty} x(t) ^2dt$
Using Periodograms	$\frac{1}{N} \sum_{k=0}^{N-1} P_{xx}[k] = \frac{1}{N} \sum_{n=0}^{N-1} x[n] ^2$	$T_{s}\sum_{k=0}^{N-1} P_{xx}[k] = T_{s}\sum_{n=0}^{N-1} x[n] ^{2}$
	Average Power	Total Energy
Periodograms	$P_{xx}[k] = \sum_{k=0}^{N-1} \frac{1}{N} X[k] ^2$	$P_{xx}[k] = \sum_{k=0}^{N-1} \frac{1}{N} X[k] ^2$
Approximated PSD & ESD	$PSD[k] = \frac{1}{N^2} X[k] ^2$	$ESD[k] = \frac{T_s}{N} X[k] ^2$
Average Power & Total Energy	$\sum_{k=0}^{N-1} PSD[k] = \sum_{k=0}^{N-1} \frac{1}{N^2} X[k] ^2$	$\sum_{k=0}^{N-1} ESD[k] = \sum_{k=0}^{N-1} \frac{T_s}{N} X[k] ^2$

Using Periodograms

	Periodic Signals	Aperiodic Signals
Two Sided F.S. Coefficient	$\frac{1}{N}X[k] = C_k$	$\frac{T_0}{N} X[k] = X(jk\omega_0)$
Approximated PSD & ESD	$\frac{ X[k] ^2}{N^2}$	$\frac{T_s}{N} X[k] ^2$
	Averaging the periodogram	Integrating the periodogram
Using Periodograms	$\frac{1}{N}\sum_{k=0}^{N-1} P_{xx}[k] = \frac{1}{N}\sum_{n=0}^{N-1} x[n] ^2$	$T_{s}\sum_{k=0}^{N-1} P_{xx}[k] = T_{s}\sum_{n=0}^{N-1} x[n] ^{2}$
	Average Power	Total Energy
Approximated PSD & ESD	$\sum_{k=0}^{N-1} PSD[k] = \sum_{k=0}^{N-1} \frac{1}{N^2} X[k] ^2$	$\sum_{k=0}^{N-1} ESD[k] = \sum_{k=0}^{N-1} \frac{T_s}{N} X[k] ^2$

FS Coefficients of Periodic and Aperiodic Signals





FS Coefficients of Periodic and Aperiodic Signals

	Periodic Signals	Aperiodic Signals
Frequency Spacing	$\Delta f = \frac{1}{N \Delta t}$	$\Delta f = \frac{1}{N \Delta t}$
Two Sided F.S. Coefficient	$\frac{1}{N}X(k)$	$\frac{\Delta t}{N} X(k)$
One Sided F.S. Coefficient	$\frac{1}{N}X(k) k=0, \ \frac{N}{2}$ $\frac{2}{N}X(k) k=1,\cdots,\frac{N}{2}-1$	$\frac{\Delta t}{N} X(k) \qquad k=0, \ \frac{N}{2}$ $\frac{2\Delta t}{N} X(k) \qquad k=1, \cdots, \frac{N}{2}-1$
Frequency Bin	$k \Delta f$	$k \Delta f$
	Average Power	Total Energy



Power Spectrum and Power Spectral Density



$$\frac{1}{N} \sum_{n=0}^{N-1} x^{2}[n] = \frac{1}{N^{2}} \sum_{k=0}^{N-1} |X[k]|^{2}$$

$$\Rightarrow \sum_{k=0}^{N-1} S[k] \Delta f$$

$$= \frac{1}{NT_{s}} \sum_{k=0}^{N-1} S[k] = \frac{1}{N^{2}} \sum_{k=0}^{N-1} |X[k]|^{2}$$

$$S[k] = \frac{T_{s}}{N} |X[k]|^{2}$$



Periodogram → Power Spectral Density

Spectrum Representation (2B)

FS Coefficients of Random Signals

Random Signals

Frequency Spacing

$$\Delta f = \frac{1}{N \Delta t}$$

Two Sided Power Spectral Density

One Sided Power Spectral Density

$P = \sum_{k=1}^{N-1} S(k) \Delta$	f

$$P = \sum_{k=0}^{N/2} S_{1}(k) \Delta f$$

$$S_{1}(k) = 2S(k) \quad k = 0, \ \frac{N}{2}$$

$$S_{1}(k) = S(k) \quad k = 1, \cdots, \frac{N}{2} - 1$$

 $\frac{1}{N\Delta t} \sum x^2 \Delta t$ $\sum S \Delta f = \frac{1}{N\Delta t} \sum S$

$$S(k) = \frac{\Delta t}{N} |X(k)|^2$$

Frequency Bin

Representation (2B)

Spectrum

 $k\Delta f$

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Power Spectrum using FFT

Discrete Fourier <u>Transform</u>

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j(2\pi/N)kn}$$
$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{+j(2\pi/N)kn}$$

$$x(t) = \sum_{k=-\infty}^{+\infty} C_k e^{+jk\omega_0 t}$$
 CTFS

$$x_{FS}(t) = \sum_{k=-M}^{+M} \gamma_k e^{+jk\omega_0 t}$$
 DTFS

 $C_k \approx \gamma_k = \frac{X[k]}{N}$ Approximated Fourier Coefficients

 $|C_k|^2 \approx \frac{|X[k]|^2}{N^2}$ Approximated Power Spectrum X = fft(x)

$$x = ifft(X)$$

Approximated Fourier Series Coefficients

fc = fft(x)/N = X/N

x = ifft(fc)*N

Spectrum Representation (2B)

Periodogram using FFT

RMS in continuous time $C_k \approx \gamma_k = \frac{X[k]}{N}$ Approximated Fourier Coefficients $|C_k|^2 \approx \frac{|X[k]|^2}{N^2}$ Approximated Power Spectrum $\frac{1}{T}\int g^2(t) dt$ $\frac{1}{N} \sum_{n=0}^{N-1} x^{2}[n] = \frac{1}{N^{2}} \sum_{k=0}^{N-1} |X[k]|^{2}$ Average Power $\left| \sqrt{\frac{\sum_{k=0}^{N-1} |X[k]|^2}{N}} \right|^2 \quad \text{RMS of sq root} \\ \text{Periodogram}$ **RMS** in discrete time $\frac{|X[k]|^2}{N} \quad k=0,1,\ldots,N-1 \quad \begin{array}{c} \text{Approximated} \\ \text{Periodogram} \end{array}$ $N \Delta$ Λ $\frac{|X[k]|}{\sqrt{N}} \quad k=0,1,\ldots,N-1 \quad \begin{array}{c} \text{Square root} \\ \text{Periodogram} \end{array}$ $\frac{1}{N\Lambda} \sum_{k=0}^{N-1} |g[k]|^2 \Delta = \frac{1}{N} \sum_{k=0}^{N-1} |g[k]|^2$

From CTFS to CTFT

Continuous Time Fourier <u>Series</u>

$$C_{k} = \frac{1}{T} \int_{0}^{T} x(t) e^{-jk\omega_{0}t} dt \qquad \Longrightarrow \qquad x(t) = \sum_{n=0}^{\infty} C_{k} e^{+jk\omega_{0}t}$$

$$C_{k} = \frac{1}{T_{0}} \int_{-T_{0}/2}^{+T_{0}/2} x_{T_{0}}(t) e^{-jk\omega_{0}t} dt \qquad \qquad x_{T_{0}}(t) = \sum_{n=0}^{\infty} C_{k} e^{+jk\omega_{0}t} \cdot \frac{2\pi}{2\pi} \cdot \frac{T_{0}}{T_{0}}$$

$$C_{k}T_{0} = \int_{-T_{0}/2}^{+T_{0}/2} x_{T_{0}}(t) e^{-jk\omega_{0}t} dt \qquad \qquad x_{T_{0}}(t) = \frac{1}{2\pi} \sum_{k=0}^{\infty} C_{k}T_{0} e^{+jk\omega_{0}t} \cdot \frac{2\pi}{T_{0}}$$

$$T_0 \rightarrow \infty$$
 $\omega_0 = \frac{2\pi}{T_0} \rightarrow d \omega$ $C_k T_0 \rightarrow X(j \omega)$ $x_{T_0} \rightarrow x(t)$

Continuous Time Fourier <u>Transform</u>

$$X(\mathbf{j}\omega) = \int_{-\infty}^{+\infty} x(t) e^{-\mathbf{j}\omega t} dt \quad \longleftrightarrow \quad x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(\mathbf{j}\omega) e^{+\mathbf{j}\omega t} d\omega$$

Spectrum Representation (2B)

CTFS and CTFT

Discrete Fourier <u>Transform</u>

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j(2\pi/N)kn}$$

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{+j(2\pi/N)kn}$$

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{+j(2\pi/N)kn}$$

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{-j(2\pi/N)kn}$$

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} x[n] e^{-j(2\pi/N)kn}$$

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} x[n] e^{-j(2\pi/N)kn}$$

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} x[n] e^{-j(2\pi/N)kn}$$

 $C_k \approx \gamma_k = \frac{X[k]}{N}$ Approximated Fourier Coefficients

Continuous Time Fourier <u>Series</u>

$$\frac{C_k T_0}{C_k T_0} = \int_{-T_0/2}^{+T_0/2} x_{T_0}(t) e^{-jk\omega_0 t} dt$$

$$\frac{X(j\omega)}{X(j\omega)} = \int_{-\infty}^{+\infty} x(t) e^{-j\omega t} dt$$

$$T_0 \rightarrow \infty$$
, $\omega_0 \rightarrow 0 \ (\omega_0 \rightarrow d \ \omega)$

$$x_{T_0}(t) = \frac{1}{2\pi} \sum_{k=0}^{\infty} C_k T_0 e^{+jk\omega_0 t} \cdot \frac{2\pi}{T_0}$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) e^{+j\omega t} d\omega$$

Spectrum Representation (2B)

Signals without discontinuity Signals with discontinuity

Sampling frequency is not an integer multiple of the FFT length

Leakage

$\begin{bmatrix} \mathbf{0}, & \frac{f_s}{2} \end{bmatrix}$



Laplace Transform $(t) e^{-st} = f(t)e^{-(\sigma + j\omega)t}$

Linear Time Domain Analysis Initial Condition



z Transform

 $f[n] z^{-n}$

Discrete Time System Difference Equation

$$z = e^{sT} = e^{\sigma T} e^{j\omega T}$$





Spectrum Representation (2B)



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